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Theoretical and experimental study of gradient-helicoid metamaterial

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1 **Abstract**

2 Resonance based acoustic metamaterials and metasurfaces exhibit outstanding phase
3 modulation capabilities with high transmission coefficient, but it is rather confined to the
4 neighborhood of fundamental or harmonic resonant frequencies. Based on the
5 impedance-matched gradient material, we theoretically and experimentally study a gradient
6 index helical-structured metamaterial that can modulate sound wavefront with near perfect
7 transmission over a broad bandwidth. Taking advantage of the linear spectral phase shifting, a
8 thin flat metalens is constructed to demonstrate sound focusing with the extraordinary
9 transmission for more than 1/3 octave band. It is anticipated that this engineered material may
10 be used in ultrasonography, ultrasound surgery, and DNA fragmentation. Moreover, with the
11 flexible adjustment of effective parameters, it can be an ideal alternative to the anisotropic
12 inhomogeneous medium and has much better stiffness.

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15 **Keywords**

16 acoustic metamaterial, helical structure, gradient index, extraordinary transmission,
17 meta-lens, focusing

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1 **1. Introduction**

2 The past few years witnessed the rapid expansion of study in the field of acoustic
3 metasurface [1–13]. It enriches the acoustic wave control methodology, offering abundant
4 unique wavefront modulation functionalities in different applications such as sound focusing
5 [1,2,12], ultrathin absorbers [7], Bessel beam [8], self-accelerating beam, anomalous
6 refraction [14] or reflection [3,4,11,15], cloaking [16,17], and so forth. Among them, most are
7 based on the generalized Snell’s law [5,16], suffering from limited bandwidth. There are two
8 reasons. First, for those approaches that utilize cavity resonance based structures [12,13,18] to
9 generate phase delay, they only work near the resonant frequencies. Secondly, if the high
10 index is introduced by coiling up space [2,9,14], the designed metasurfaces are always
11 accompanied by the impedance mismatching and the Fabry–Pérot resonances incurred. The
12 dramatical decrease of the transmission coefficient occurs near to the Fabry–Pérot resonant
13 frequencies. To extend the working frequency range, corrugated surface [19,20] and
14 circular-holed cube [21] were adopted. They can effectively modulate the reflected acoustic
15 wave, but not suitable for the transmitted wave. In this context, metamaterials with varying
16 geometrical shape [10,22–26] were introduced to improve the spectral performance, as the
17 graded interface between metamaterial and background media can decrease the reflection
18 with a better impedance matching behavior. For some of the design [22], the high index
19 region is restricted by the viscous or structural loss because of the structural complexity. Also,
20 many approaches [24,27] were in a primary stage, and more experimental investigation is
21 needed. Notably, by ‘compressing the space’ by using helicoid surface [28], a broadband
22 dispersion-free metamaterial can be obtained, and its effective parameters can be adjusted
23 precisely by changing the helical pitch. Former works attempted to make an
24 impedance-matched layer by using gradient pitch distribution [27,29–31] and sound devices
25 with diverse functionalities has been demonstrated with numerical results. However, little
26 systematic investigation has been conducted on the overall reconstruction of the uniform
27 helicity design, especially the theoretical analysis on subsequent benefits in the frequency
28 response of phase delay and transmission efficiency. Also, there lacks the experimental work
29 to evaluate the effect of viscous and thermal loss and demonstrate the broadband sound
30 wavefront modulation with high energy transitive ratio.

31 In this study, we adopt helical-structured elements whose pitch is designed to change
32 linearly along the middle axis of the cylindrical unit cells, matching the impedance of air with
33 small helicity density at both ends. By using the transfer matrix model, we have theoretically
34 verified the validity of the effective media model for graded helicoid metamaterial. We have
35 measured the performance of the energy transmission efficiency and evaluated the dispersion
36 and loss property of a helicoid unit cell. We have proposed a flat focal lens working in a wide
37 frequency range and experimentally measure the sound field of such a metalens. The graded
38 helicoid metamaterial has been thoroughly studied theoretically and experimentally. Make it

1 possible to design metamaterials whose effective characteristics and behaviors can match a
2 corresponding inhomogeneous media. Moreover, the acoustic behavior can be predicted by
3 the effective parameter model, which paves the way for “inverse design” from demanded
4 character to a corresponding metamaterial structure. The helicoid bi-anisotropic metalens
5 enriched the flexibility of acoustic metasurface.

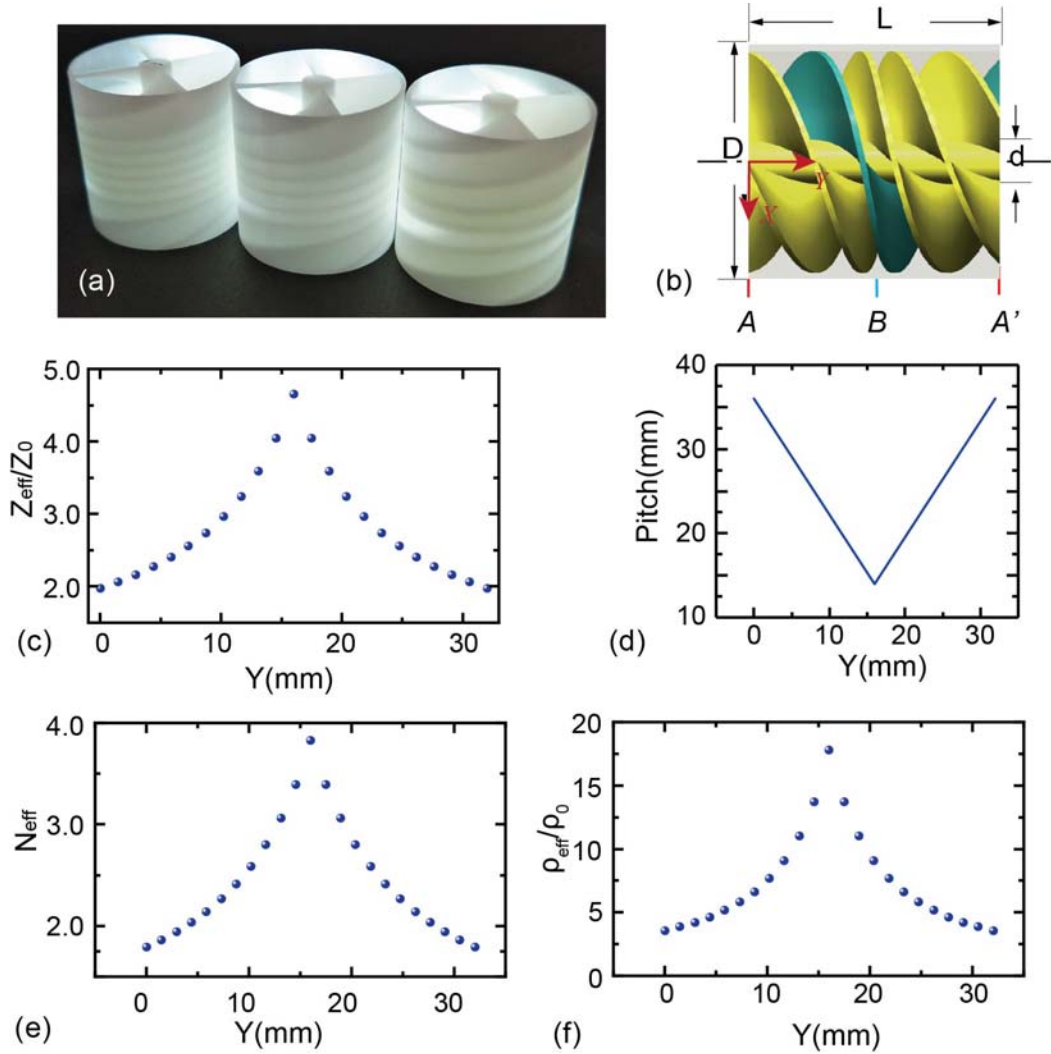
6 In section 2, we study the effective parameter of the helicoid metamaterial and calculate its
7 performance over spectrum in theory. In section 3, we experimentally test the performance of
8 a single element of the metamaterial both in frequency domain and time domain. Also, the
9 loss and dispersion properties are evaluated by experiment. In section 4, to show the
10 application of the gradient helicoid metamaterial, we design a broadband focal lens and verify
11 its performance with simulation and experimental measurement. In the end, we present the
12 conclusions and appendices.

13 **2. The effective model of the helicoid element**

14 In this section, we demonstrate the characteristics and performance of the proposed
15 metamaterial unit cell. A theoretical model of the inhomogeneous medium has been
16 conducted to calculate the phase and transmission spectra of the element.

17 **2.1 The structure and its acoustic characteristics**

18 The performance of the helicoid acoustic metamaterial with constant helical pitch decays
19 away from the Fabry–Pérot resonance frequencies, due to the impedance mismatch at two
20 ends of the unit cells. To investigate the behavior of the graded helicoid metamaterial over a
21 broad bandwidth and offer new wavefront modulation capability, we propose a design with
22 linearly pitch distribution, by which the impedance matching has been improved on the
23 interfaces. As shown in Fig. 1(a), the unit element of the proposed helical structure has the
24 largest pitch at two ends and the smallest pitch in the middle, symmetric about the middle
25 point. The effective impedance of such gradient helical structured unit cells is no longer
26 uniform along the wave propagation but instead decided by the on-site pitch value.



1

2 Fig. 1. The unit cell of the graded helicoid metamaterial. (a) Unit cells' photograph. (b)
 3 Detailed interior design of helical unit cell with four blades and gradient pitch. In this study,
 4 the inner diameter, outer diameter, and total length are d (6 mm), D (30 mm), and L (32 mm),
 5 respectively. (c)-(f) The distribution of effective impedance, pitch, effective refractive index
 6 and effective mass density of the gradient-pitched element.

7 The analytical analysis of this helical-structured metamaterial with gradient-pitched
 8 element design can be resolved with the effective medium model. We obtain the effective
 9 refractive index, mass density and impedance corresponding to a certain pitch through the
 10 retrieving method which is based on the transmission and reflection spectrum [32]. For an
 11 element with a constant diameter and gradient pitch illustrated in by Fig. 1(d), the effective
 12 parameter variations along the wave propagation direction, caused by different pitch values
 13 are calculated and plotted in Fig. 1(c)-(e). It is obvious that all three parameters share the
 14 same variation trend that is against the pitch change. From the end to the middle point along
 15 the Y-axis, they all increase gradually with decreasing pitch value and a tighter arrangement
 16 of the blades.

2.2 The performance over the spectrum

For a helical duct with a constant diameter, the effective parameter can be the function of the pitch. In this case, by selecting appropriate pitch distribution, a flexible and ideal inhomogeneous media can be obtained for diverse functionalities and working conditions. Based on this, the behavior of the helical-structured metamaterial elements can be estimated by the effective parameters instead of considering the circuitous geometric space. To analytically describe the relationship between the effective parameter and the position along the Y-axis, an inversely proportional function is put into use in table 1 and 2 to fit the distribution of index and mass density of the cell displayed in Fig. 1(d). As a symmetric profile with respect to the middle of the unit cell, the effective materials can be divided into two layers corresponding to the increasing (layer II in Fig. 2) and decreasing (layer III in Fig. 2) trend of parameters respectively. After the above operation, the geometrical model of the graded helicoid metamaterial has been replaced by the effective medium.

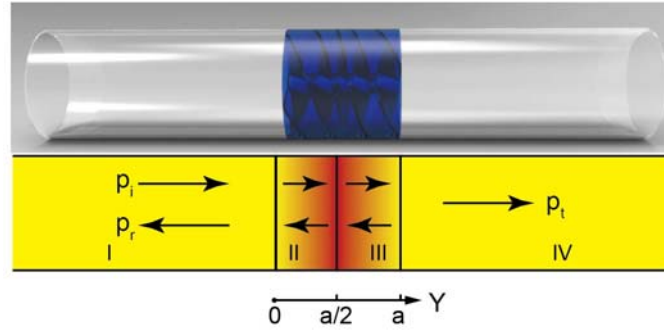


Fig. 2. The illustration of a gradient-helical-structured unit cell and background's acoustic indices

Table 1. The fitting of the effective refractive index

Interval of y	$0 - \frac{L}{2}$	$\frac{L}{2} - L$	
Expression	$N_{eff} = \frac{1}{\alpha_1 y + \beta_1}$	$N_{eff} = \frac{1}{\alpha_2 y + \beta_2}$	
Independent variable	y		
Dependent variable	N_{eff}		
Method	Trust Region method [33]		
Coefficients to be fitted	α_1	β_1	α_2 β_2
Result	-19.66 m^{-1}	0.5804	19.66 m^{-1} -0.04862
Sum of squares due to error	0.04375		0.04375

1 **Table 2.** The fitting of the effective mass density

Interval of y	$0 - \frac{L}{2}$	$\frac{L}{2} - L$		
Expression	$\rho_{eff} = \frac{1}{\mu_1 y + \nu_1}$	$\rho_{eff} = \frac{1}{\mu_2 y + \nu_2}$		
Independent variable	y			
Dependent variable	ρ_{eff}			
Method	Trust Region method			
Coefficients to be fitted	μ_1	ν_1	μ_2	ν_2
Result	-10.98 $\text{kg}^{-1} \text{m}^2$	0.2218 $\text{kg}^{-1} \text{m}^3$	10.98 $\text{kg}^{-1} \text{m}^2$	-0.1295 $\text{kg}^{-1} \text{m}^3$
Sum of squares due to error	0.7205 $\text{kg}^2 \text{m}^{-6}$		0.7205 $\text{kg}^2 \text{m}^{-6}$	

2 Outside the metamaterial region, the one-dimensional monochromatic wave equation in the
3 background medium (layers I and IV) is (time dependency $e^{j\omega t}$ has been eliminated [34])

$$4 \quad \frac{d^2 p(y)}{dy^2} + \left(\frac{\omega}{c}\right)^2 p(y) = 0 \quad (1)$$

5 with the solution being expressed as

$$6 \quad p(x) = \Psi_I^1 e^{-jky} + \Psi_I^2 e^{jky} \quad (2)$$

7 where Ψ_I^1 and Ψ_I^2 are constants which can be determined by the boundary conditions.

8 For the particle velocity along the propagation direction, we have

$$9 \quad \rho \frac{\partial v}{\partial t} = -\frac{\partial p}{\partial y} \quad (3)$$

10

11 and thus

$$12 \quad v_I(x) = \Psi_I^1 \frac{ke^{-jky}}{\omega\rho_0} - \Psi_I^2 \frac{ke^{jky}}{\omega\rho_0} \quad (4)$$

13 The wave equation of the inhomogeneous media in layers II and III has the form below [35]:

$$14 \quad \frac{d^2 p_1(y)}{dy^2} + \left(\frac{\omega}{c} n_{eff}(y)\right)^2 p_1(y) - \left(\frac{d \ln \rho_{eff}(y)}{dy}\right) \frac{d}{dy} p_1(y) = 0 \quad (5)$$

15 where $n_{eff}(y)$ and $\rho_{eff}(y)$ are the space-dependent variables that can be expressed as

$$n_{eff} = \begin{cases} \frac{1}{\alpha_1 y + \beta_1}, 0 \leq y \leq L/2 \\ \frac{1}{\alpha_2 y + \beta_2}, L/2 < y \leq L \end{cases}, \rho_{eff} = \begin{cases} \frac{1}{\mu_1 y + \nu_1}, 0 \leq y \leq L/2 \\ \frac{1}{\mu_2 y + \nu_2}, L/2 < y \leq L \end{cases}, \quad (6)$$

where α_i , β_i , μ_i and ν_i ($i=1,2$) are the coefficients obtained by curve fitting in table 1 and table 2.

The solution of Eq. (5) has the form

$$p_{II}(x) = \Psi_{II}^1 \Lambda(y) + \Psi_{II}^2 \Theta(y) \quad (7)$$

where Ψ_{II}^1 and Ψ_{II}^2 are constants for the two linearly independent solutions denoted by $\Lambda(y)$ and $\Theta(y)$. The details of the solution are presented in Appendix B. Combine Eq. (3) and (7), the particle velocity in layer II is

$$v_{II}(y) = \Psi_{II}^1 \frac{j}{\omega \rho} \frac{d\Lambda(y)}{dy} + \Psi_{II}^2 \frac{j}{\omega \rho} \frac{d\Theta(y)}{dy} \quad (8)$$

In layer III the sound pressure and particle velocity have the same form as that in layer II but the values of α , β , ν and μ are different. The independent solution can be denoted by $\Lambda'(y)$ and $\Theta'(y)$. With the continuity of sound pressure and normal particle velocity at the interfaces, we can obtain the following relationship for the background and effective medium layers.

$$\begin{cases} p_I(0) = p_{II}(0) \\ v_I(0) = v_{II}(0) \end{cases}, \begin{cases} p_{II}\left(\frac{a}{2}\right) = p_{III}\left(\frac{a}{2}\right) \\ v_{II}\left(\frac{a}{2}\right) = v_{III}\left(\frac{a}{2}\right) \end{cases}, \begin{cases} p_{III}(a) = p_{IV}(a) \\ v_{III}(a) = v_{IV}(a) \end{cases} \quad (9)$$

By changing Eq. (9) to matrix form and eliminating the middle term, the relationship between layer I and layer IV is

$$\mathbf{T} = \mathbf{T}_{II}|_{y=a} \cdot \mathbf{T}_{II}|_{y=\frac{a}{2}}^{-1} \cdot \mathbf{T}_I|_{y=\frac{a}{2}} \cdot \mathbf{T}_I|_{y=0}^{-1} \cdot \mathbf{M}|_{y=0} \quad (10)$$

In Eq. (10),

$$\mathbf{T}_I = \begin{pmatrix} \Lambda(y) & \Theta(y) \\ \frac{j}{\omega \rho(y)} \frac{d\Lambda(y)}{dy} & \frac{j}{\omega \rho(y)} \frac{d\Theta(y)}{dy} \end{pmatrix}, \mathbf{T}_{II} = \begin{pmatrix} \Lambda'(y) & \Theta'(y) \\ \frac{j}{\omega \rho(y)} \frac{d\Lambda'(y)}{dy} & \frac{j}{\omega \rho(y)} \frac{d\Theta'(y)}{dy} \end{pmatrix},$$

$$\mathbf{M} = \begin{pmatrix} e^{-jk_y} & e^{jk_y} \\ \frac{ke^{-jk_y}}{\omega \rho_0} & \frac{ke^{jk_y}}{\omega \rho_0} \end{pmatrix}.$$

Based on the continuity of sound pressure and normal particle velocity, the relationship

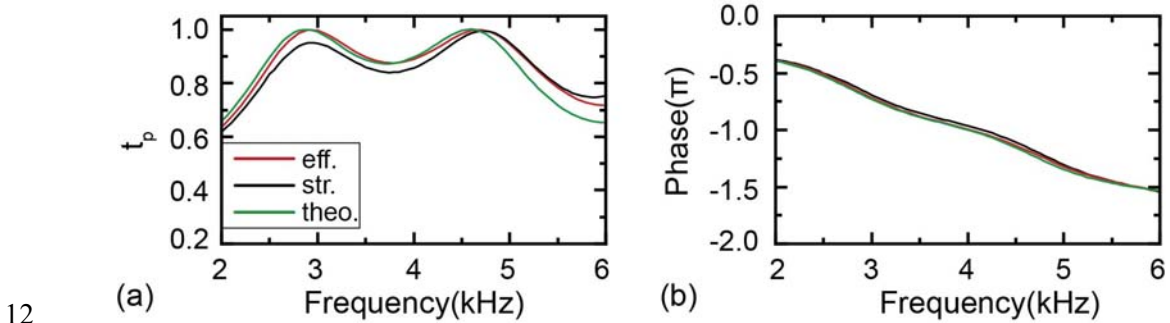
1 between the pressure p_i of incidence and transmission p_t follows the equation

$$2 \quad \mathbf{T} \begin{bmatrix} p_i \\ p_r \end{bmatrix} = p_i e^{-jkL} \begin{bmatrix} 1 \\ \frac{k}{\omega \rho_0} \end{bmatrix}, \quad \mathbf{T}^{-1} = \begin{bmatrix} t_{inv11} & t_{inv12} \\ t_{inv21} & t_{inv22} \end{bmatrix} \quad (11)$$

3 In Eq. (11), k and ω are the wave number and angular frequency of the plane wave;
4 ρ_0 is the mass density of air. In this case, the overall complex transmission coefficient of
5 sound pressure t_p is derived as

$$6 \quad t_p = \frac{e^{jkL}}{t_{inv11} + \frac{k}{\rho_0} \frac{1}{t_{inv12}}}. \quad (12)$$

7 One needs to take into consideration is the small difference between the real structure and
8 the effective model. Because of the large pitch of helicity at both ends of cells, the effective
9 model can be slightly shorter than the length of the blade region (32 mm). Thus, for the
10 numerical calculation and simulation model with an effective parameter, we cut 2 mm at
11 both ends for more accurate results.



12 (a) (b)
13 Fig. 3. The calculated performance of an element. (a) The transmission spectrum of normal
14 incident plane waves crosses the unit cell. (b) The corresponding phase spectrum response. (a)
15 and (b) share the same legend in (a). Note of legend: eff.—simulation results by using
16 effective media, str.—simulation by using the helicoid model, theo. —numerical results
17 corresponding to the theoretical analysis.

18 The calculated sound pressure transmission and phase delay through the gradient index unit
19 cell is shown in Fig. (3). The curve in Fig. 3(a) clearly shows the transmission coefficient
20 higher than 80% over the most frequency range. At the same time, the phase delay shows
21 smooth variation with a gentle slope over the studied frequency range. It is because the
22 gradient index along the propagation direction weakens the strong dispersion of phase shift
23 caused by specific Fabry-Pérot resonance. With such a smooth curve, controlling the phase
24 delay for phase engineering within the whole wide frequency range of interests is easier and
25 more practical.

26 The good agreement of three groups of results in Fig. (3) indicates the feasibility of the

1 effective medium model and the theoretical analysis. The small difference of transmission
2 coefficient of sound pressure may come from the round-off error of the numerical
3 computation and thickness of the structure's blades. As we can see in the illustration of the
4 phase delay (Fig. 3(b)), a slight shift can be neglected. On the one hand, the result shows the
5 validity of the inhomogeneous medium model to replace the real structure. Thus, the designed
6 metamaterial with air-like acoustic properties has relatively high stiffness, which can be used
7 as inhomogeneous sound media. On the other hand, the conformity guarantees the 'inverse
8 design' from required material parameters' distribution to the related helical structure. This
9 property remains a huge potential for the flexible design of acoustic metamaterial. The
10 computation above are all in a lossless condition and the influence of viscous and thermal loss
11 has been evaluated by experiment in the following section.

12 **3. Experimental investigation of the helicoid element**

13 In this section, the experiment results are illustrated to show the performance and
14 characteristics of the graded helicoid metamaterial.

15 **3.1 Experimentally tested performance**

16 The performance of the newly designed helical-structured acoustic metamaterial unit cell
17 with gradient pitch has been experimentally measured inside an impedance tube using the
18 4-microphone 2-load method, as shown in Fig. 4(a). The samples were manufactured by 3D
19 printing. The huge impedance mismatch between the photopolymer (Somos GP Plus)
20 material we used to fabricate the sample and air makes it possible for us to regard it as being
21 rigid in theory model. We performed both time and frequency related experiments. To
22 demonstrate that the new gradient-pitched helical-structured design still possesses the sound
23 deceleration function, a short pulse was generated and sent inside the impedance tube. The
24 recorded transmission signals with and without the metamaterial sample are shown in Fig.
25 4(b). It is evidently clear that, by adding the metamaterial unit cell, the sound signal is further
26 delay by about 1.19×10^{-4} s. The time delay functionality for the signal transmission has been
27 verified.

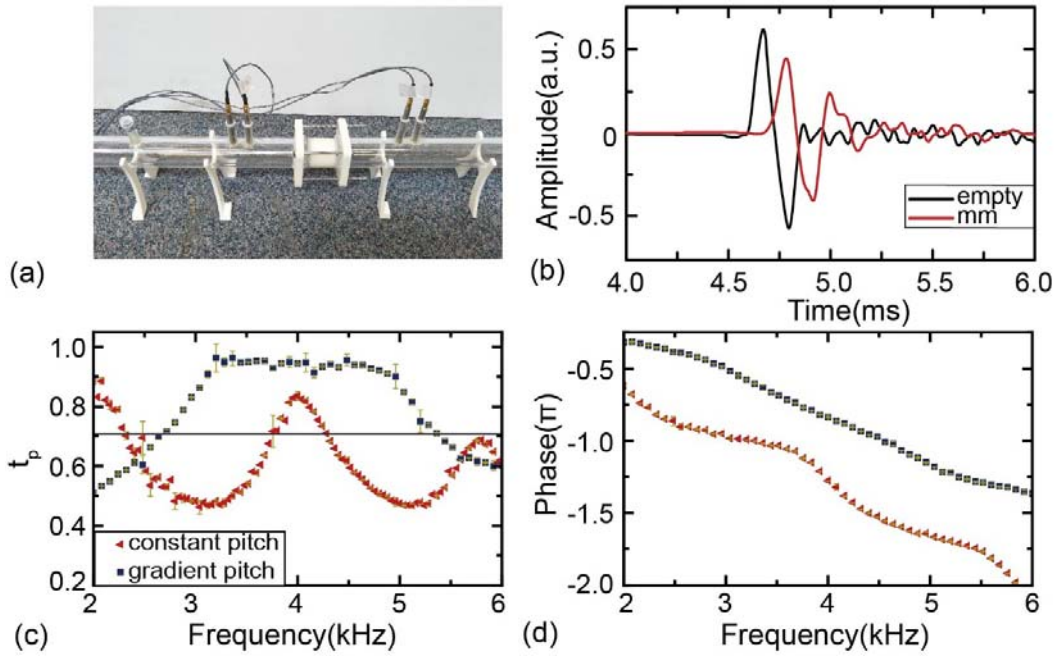


Fig. 4. Measurement of gradient helical-structured metamaterial element. (a) The setup of the lab-made impedance tube. The white-colored cell is fixed in the middle by two flanges. (b) The delay of the pulse sound signal introduced by the gradient-pitched helical-structured element. (c) The sound pressure transmission coefficient of a uniform-pitched and gradient pitched element. (d) Phase shift spectra of a uniform-pitched and gradient pitched element. For the uniform-pithed element, the leading pitch is 20 mm; for the gradient pitched unit cell, the pitch distribution is the same as Fig. 1(d). Note of legend: mm—with metamaterial layer.

From the uniform pitch distribution to the graded model, the impedance mismatch has been weakened, so as the Fabry–Pérot resonance related to the reflection at interfaces. The difference can be observed by the comparison of transmission spectrum between elements with constant pitch and gradient pitch. Fig. 4(c) shows the experimentally measured spectral transmission performance. As anticipated, the transmission coefficient of the uniform pitched model changes violently, due to the relatively pure Fabry–Pérot resonance mode. The 3-dB half-power bandwidth is rather narrow, about 0.52 kHz. Represented by the blue curve, the transmission through the gradient-pitch unit cell remains at high values close to unity over a wide frequency range, in good agreement with the prediction from the transfer matrix method. The 3dB half-power band is now 2.6 kHz wide, covering from 2.68 kHz to 5.28 kHz, about one octave. Such extraordinary sound pressure transmission of the gradient-pitched model over spectrum can be attributed to two physical mechanisms: the better impedance matching with the background medium at two ends of the unit cell due to the larger pitch there, and the gradient index profiles that can bridge more sound energy through to counter the effects of specific Fabry–Pérot resonance mode. Fig. 4(d) demonstrates the measured phase shift of the two unit cells. While the gradient-pitched unit cell shows a smooth phase delay curve without

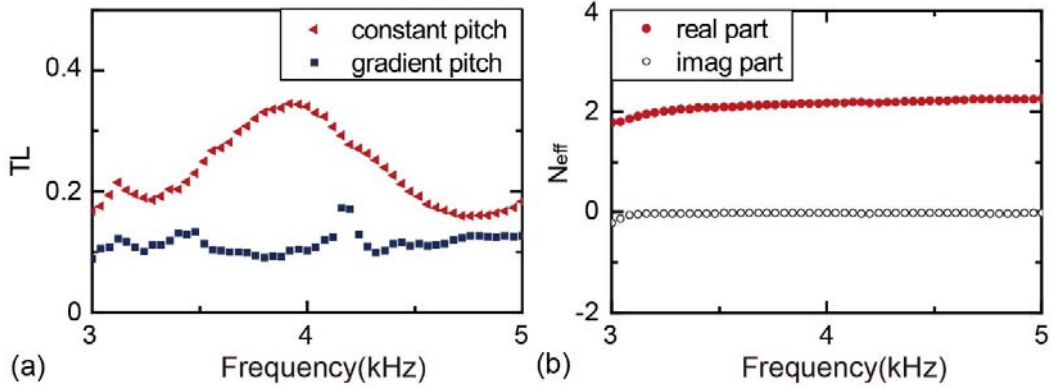
1 a notable inflection point in the 3-dB band, the phase shift offered by the uniform-pitched unit
 2 cell fluctuates around the resonant frequency at 4 kHz.

3.2 Evaluation of the loss and dispersion

4 The previous work [30] pointed out the dispersion-free property of the uniform-pitched
 5 element, while the dispersion property of the gradient-pitched model has not been
 6 experimentally studied. To sufficiently evaluate the characteristic of the helicoid metamaterial,
 7 we calculated the spectra of energy transmission and reflection. Thus, the viscous and thermal
 8 loss of the element can be obtained according to the relationship

$$9 \quad TL = 1 - t_l - r_l \quad (13)$$

10 where TL is the energy loss, t_l and r_l are the transmission and reflection coefficients of
 11 the energy. We can see that the thermal and viscous loss keeps at a low level over the
 12 spectrum (Fig. 5(a)). As a contrast, the energy loss of a uniform-pitched element shows a
 13 larger loss near the resonance frequency. The small loss can also be further demonstrated by
 14 the effective refractive index. As the gradient distribution of the helical pitch, retrieving of
 15 effective parameters cannot be straightforward. However, we can measure the average value
 16 of the effective refractive index by taking it as a homogeneous medium. The real part of the
 17 index in Fig. 5(b) shows the dispersion-free property of the graded helical duct over the
 18 spectrum. The imaginary part – the attenuation coefficient—is near to zero and keeps flat
 19 during the calculated frequency range. Hence the gradient-pitched helical structure can be
 20 described by a lossless model reasonably.



22 Fig. 5. The energy loss and effective index. (a) The measured viscous and thermal loss of
 23 uniform and graded helicoid cell. (b) The measured average refractive index of the graded
 24 helicoid cell.

25 4. A helicoid-metamaterial-based diffractive focal lens

26 In this section, a flat focal lens works in a wide frequency range has been realized.
 27 According to the characteristic of the lens, the position of the focal spot at the different
 28 operating frequency is calculated and verified by experiment.

4.1 Lens design

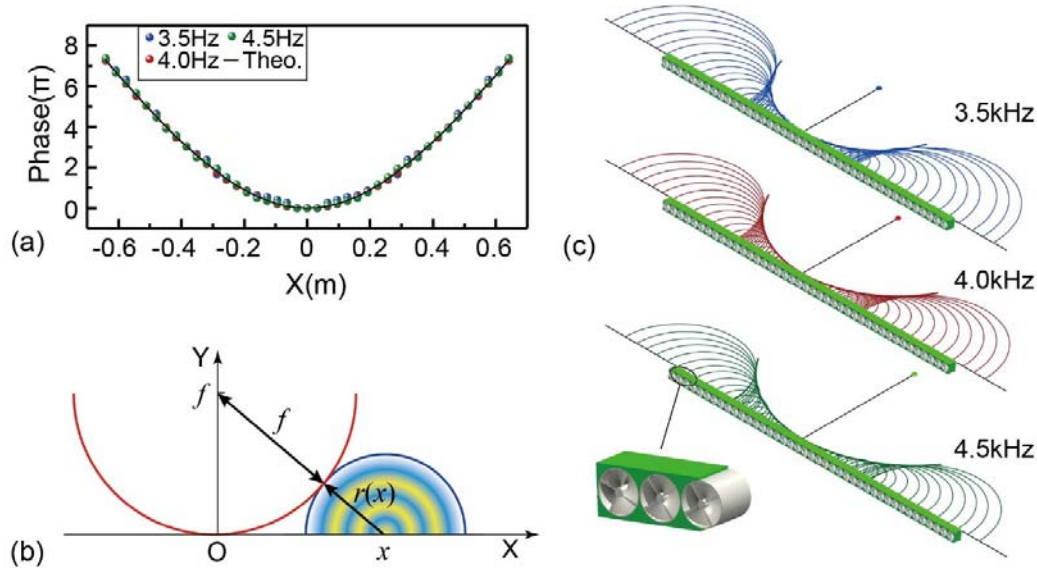
The new helical-structured acoustic metamaterial with gradient pitch offers extraordinary sound transmission and smooth phase adjustment within a wide frequency range. Therefore, it is now possible to realize a series of phase delays by appropriate pitch distributions, without the fear of suffering from the lower transmission. With the smooth phase shift, the difference of phase delay between any two unit cells keeps invariable over the spectrum. Taking advantage of this unique characteristic, we design and fabricate a thin flat focal lens with the array of new gradient-pitched helical-structured acoustic metamaterial unit cells, whose thickness is less than half of the smallest wavelength. To design a focal lens for normally incident transmitted plane wave, the geometrical relationship illustrated in Fig. 6(b) should be satisfied with the equation

$$\varphi(x) = \frac{2\pi f}{c} \left(\sqrt{x^2 + F^2} - F \right). \quad (14)$$

Eq. (14) gives the required phase adjustment distribution along the surface the of the flat focusing lens, where φ denotes the phase delay at any specific location x , f is the operating frequency, c is the sound speed in the air (343.2 m s^{-1}), and F is the focal length. This continuous phase profile is further discretized into 41 segments so that the whole lens can be constructed with 41 individual unit cells embedded in a holding frame, just as exhibited in Fig. 6(c). The required phase modulation at each of the forty-one unit cells is achieved by adjusting the leading pitch. We subsequently calculate the phase profiles for three different frequencies 3.5 kHz, 4 kHz and 4.5 kHz and present the discretization results in Fig. 6(a). Because of the smooth phase delay of the elements about frequency, any two elements can keep a relatively constant phase difference over the spectrum, which forms a uniform phase profile. This paves the way towards realizing broadband sound focusing with one set of gradient-pitched helical-structured acoustic metamaterial unit cells array. Since the consistency of the phase distribution about the frequency, the position of the focal point is frequency dependent and can be predicted. In Eq. (14), taking F as the parameter to be fitted, x and $\varphi(x)$ as the independent and dependent variables, we can obtain the focal length at different frequencies with curve fitting method. In the process of fitting, we firstly have an expression

$$y = \frac{c}{2\pi f} \varphi(x) = \sqrt{x^2 + F^2} - F \quad (15)$$

where y is the wavefront based on the phase profile. By using the trust region fitting method, the focal length F can be obtained. As shown in Fig. 6(c), the radius of the output wavefront envelope, i.e., the focal length, will increase with frequency, being 0.40 m at 3.5 kHz, 0.50 m at 4 kHz and 0.58 m at 4.5 kHz, respectively.

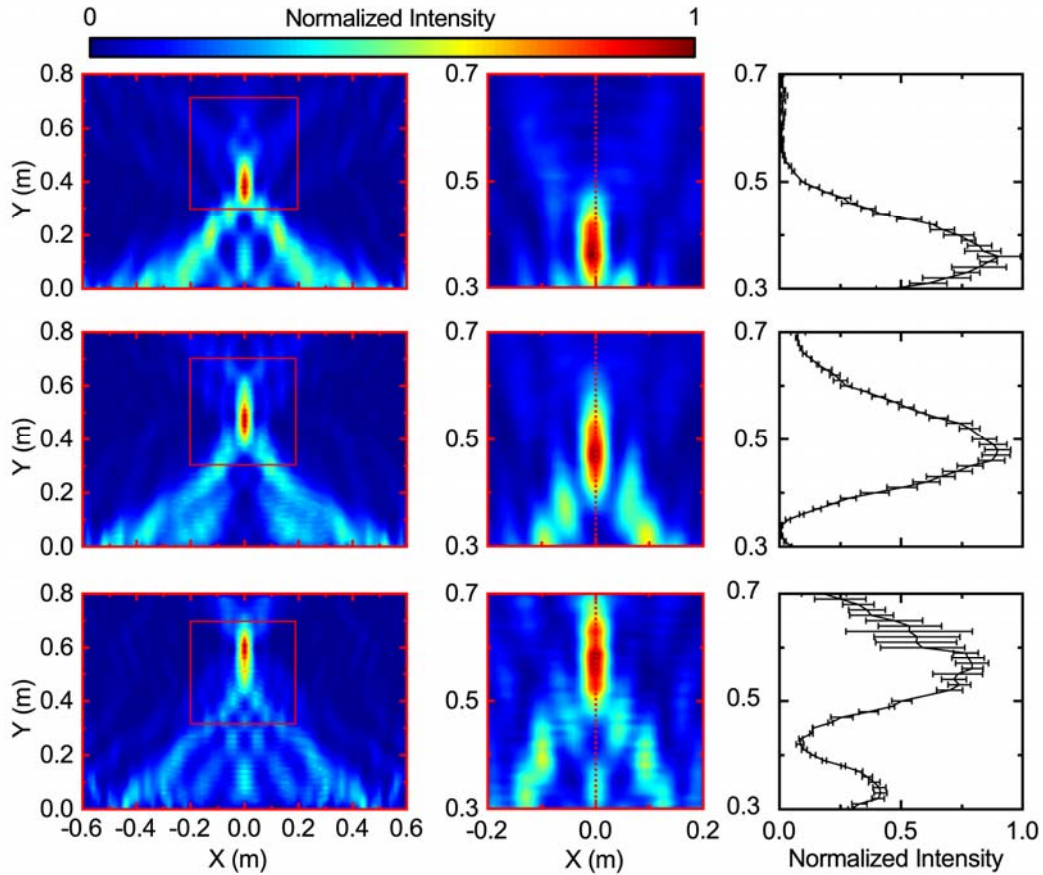


1
2 Fig. 6. Design of the focal lens with excellent sound transmission. (a) Phase profile
3 requirement of the flat focal lens for different sound frequencies. The dots represent the phase
4 of each unit cell at the corresponding positions, and the black line is the theoretically designed
5 phase profile at 4 kHz by Eq. (14). (b) The geometric relationship between focal length and
6 the wavefront of a point source at position x . The gradual change in radiation means a unit
7 cell can be regarded as a point source. (c) The focal point for three operating frequencies. The
8 semicircle is the wavefront from each unit cell. The envelope curve obtained by fitting
9 method appears in bold is the focusing wavefront. At the bottom left corner of (c), the
10 structure of the flat lens is magnified partially. Forty-one cells are embedded in the holey
11 frame whose total width is 1.312 m.

12 4.2 Field measurement of broadband focusing

13 We conducted simulation and experiments to validate the thin flat focal lens design. The
14 sound intensity patterns for three separated frequencies 3.5 kHz, 4 kHz and 4.5 kHz are
15 extracted and plotted in Fig. 7. Due to the limited range of our sound mapping system, we
16 only measured the region around the focal point, marked with red boxes in Fig. 7(a)-(c) and
17 illustrated by Fig 7(d)-(f). From the well-matched simulation and experiment results, it is
18 evidently clear that the thin flat lens can focus incident acoustic waves of different
19 frequencies covering 1/3 octave coverage. The shape of the focal spot is clear for all three
20 frequencies. The sound intensity at the focal point also remains on the same level for three
21 frequencies, indicating similar high transmission over the spectrum. Simulations for incident
22 acoustic waves with frequencies different from the previous three ones were also conducted.
23 In all cases, the similar extraordinary sound transmission and focusing effects can be clearly
24 observed. Please refer to Appendix C for the predicted behavior and the pressure field. The
25 high energy transmission coefficient in the whole operating frequency range is witnessed by
26 almost the same maximum intensity under 3.5 kHz, 4 kHz and 4.5 kHz as is shown in Fig.

1 7(g)-(i). What is noteworthy is that in the wave travelling direction the intensity drops sharply
 2 after passing through the focal spot. This behavior is of great significance in contributing to
 3 the concentration of energy. By observing the position of the peak, the experimentally
 4 determined focal length is 0.36 m, 0.47 m, and 0.56 m for the corresponding frequency in Fig.
 5 7(g)-(i). The small difference between the theoretical and experimental results of focal length
 6 indicates the good controllability and high accuracy of the lens. The error of the experiment
 7 may come from the slight shaking of the microphone during the test process. The intensity
 8 gradient near the focal spot is so steep that a slight shaking may lead to a large error.



9

10 Fig. 7. The intensity field of broadband sound focusing. (a)-(c) The simulated sound intensity
 11 field distribution at 3.5 kHz, 4 kHz and 4.5 kHz, respectively. The red blocks inside mark the
 12 scanning area adopted in the experiments. (d), (e) and (f) are the experimentally measured
 13 sound intensity field distribution at 3.5 kHz, 4 kHz and 4.5 kHz, respectively. (g), (h) and (i)
 14 are the transverse cross-section intensity distribution along the Y-axis (the red dash lines) at
 15 $x=0$ for 3.5 kHz, 4 kHz and 4.5 kHz, respectively.

16 Conclusions

17 In this study, we investigated the acoustic characteristic of the graded helicoid metamaterial
 18 and constructed a broadband focal lens with the studied element cell. With the gradient pitch
 19 design, the new helical-structured acoustic metamaterial provides better impedance matching
 20 with the background medium at two ends and bridges more sound energy through to counter

1 the effects of Fabry–Pérot resonance mode. A more detailed theoretical model has been
 2 established for the evaluation of the unit cell’s performance and providing the possibility of
 3 inverse design from acoustic parameter distribution to its geometric parameter counterparts.

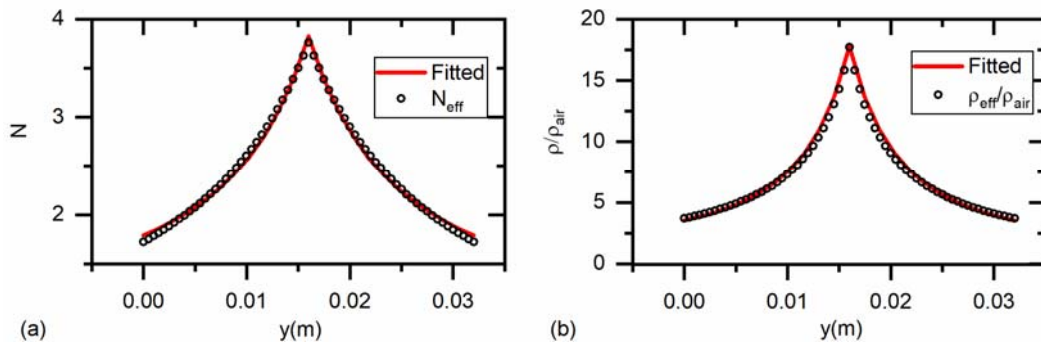
4 We experimentally verified its performance and the validity of the theoretical analysis. The
 5 measured effective index shows the low viscous and thermal loss over the spectrum and the
 6 dispersion-free property of the graded helicoid metamaterial, adding substantial evidence for
 7 the practical use. We subsequently design and fabricate a thin meta-lens with the
 8 gradient-pitched helical-structured metamaterial and the frequency-dependent focal point
 9 position has been calculated. The high performance of sound focusing effect is then verified
 10 experimentally with the lens over multiple frequencies. This study provides a flexible and
 11 precise alternative for the inhomogeneous media. The helical structures have adjustable
 12 acoustic parameters and stiff body. It will not only benefit the physics research on realizing
 13 wavefront modulation with novel passive materials but also be anticipated that this new
 14 metamaterial might be used in ultrasonography, ultrasound surgery or DNA fragmentation,
 15 which need to focalize the energy to a specific region. By using the gradient helicoid
 16 metamaterial to meet the requirement, we can make gradient index to obtain extremely high
 17 sound pressure through wave compression, which can work in a wide frequency range.

18 Acknowledgements

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 21 G-UA8T] from the Hong Kong Polytechnic University.

22 Appendix A. The fitted results of effective parameters

23 Firstly, the trust region method is commonly used for curve fitting or optimization, Helfrich
 24 [33] gave a detailed description of curve fitting. Secondly, to show the excellent agreement
 25 between the fitting result and the original data, we put a chart for comparison. The excellent
 26 agreement between the fitting result and effective parameter shows the efficiency of the
 27 method.



28 (a) The comparison of the fitted index and the
 29 original data (b) The comparison of the fitted density and the original data

1 Appendix B. The solution of the inhomogeneous equation

2 Considering the complexity of the mentioned inhomogeneous equation, we resort to the
3 symbolic calculation of computer software. The solution of Eq. (5) can be written as

$$\begin{aligned}
 p_{II}(x) &= \Psi_{II}^1 F([n_1, n_2], d, z) \cdot (\alpha y + \beta)^{-\frac{1}{2} \frac{\sqrt{\alpha^2 c^2 - 4\omega^2} - c\alpha}{c\alpha}} + \\
 &\Psi_{II}^2 F([n_1', n_2'], d', z') \cdot (\alpha y + \beta)^{-\frac{1}{2} \frac{\sqrt{\alpha^2 c^2 - 4\omega^2} + c\alpha}{c\alpha}} \\
 &= \Psi_{II}^1 \Lambda(y) + \Psi_{II}^2 \Theta(y)
 \end{aligned} \tag{B.1}$$

5 where $F([n_1, n_2], d, z)$ is the hypergeometric function which has the form

$$F([n_1, n_2], d, z) = \sum_{k=0}^{\infty} \frac{z^k \Gamma(n_1 + k) \Gamma(n_2 + k) \Gamma(d)}{k! \Gamma(n_1) \Gamma(n_2) \Gamma(d + k)}. \tag{B.2}$$

7 Meanwhile, Ψ_{II}^1 and Ψ_{II}^2 are constants for the two linearly independent solutions denoted
8 by $\Lambda(x)$ and $\Theta(x)$. For details,

$$\begin{aligned}
 n_1 &= -\frac{1}{2} \frac{\sqrt{\alpha^2 c^2 - 4\omega^2} - 2j\omega - c\alpha}{c\alpha}, n_2 = -\frac{1}{2} \frac{\sqrt{\alpha^2 c^2 - 4\omega^2} + 2j\omega - c\alpha}{c\alpha} \\
 d &= \frac{-\sqrt{\alpha^2 c^2 - 4\omega^2} + c\alpha}{c\alpha}, z = \frac{(\alpha y + \beta)\mu}{-\alpha\nu + \beta\mu} \\
 n_1' &= -\frac{1}{2} \frac{\sqrt{\alpha^2 c^2 - 4\omega^2} - 2j\omega + c\alpha}{c\alpha}, n_2' = -\frac{1}{2} \frac{\sqrt{\alpha^2 c^2 - 4\omega^2} + 2j\omega + c\alpha}{c\alpha} \\
 d' &= \frac{\sqrt{\alpha^2 c^2 - 4\omega^2} + c\alpha}{c\alpha}, z' = \frac{(\alpha y + \beta)\mu}{-\alpha\nu + \beta\mu}
 \end{aligned}$$

10 Following the property of hypergeometric functions [36], as two general solutions of the
11 inhomogeneous equation, $\Lambda(y)$ and $\Theta(y)$ are linearly independent.

12 To ensure correctness, we put $\Lambda(y)$ and $\Theta(y)$ back in to

$$D = \frac{d^2 p_1(y)}{dy^2} + \left(\frac{\omega}{c} n_{eff}(y) \right)^2 p_1(y) - \left(\frac{d \ln \rho_{eff}(y)}{dy} \right) \frac{d}{dy} p_1(y) \tag{B.3}$$

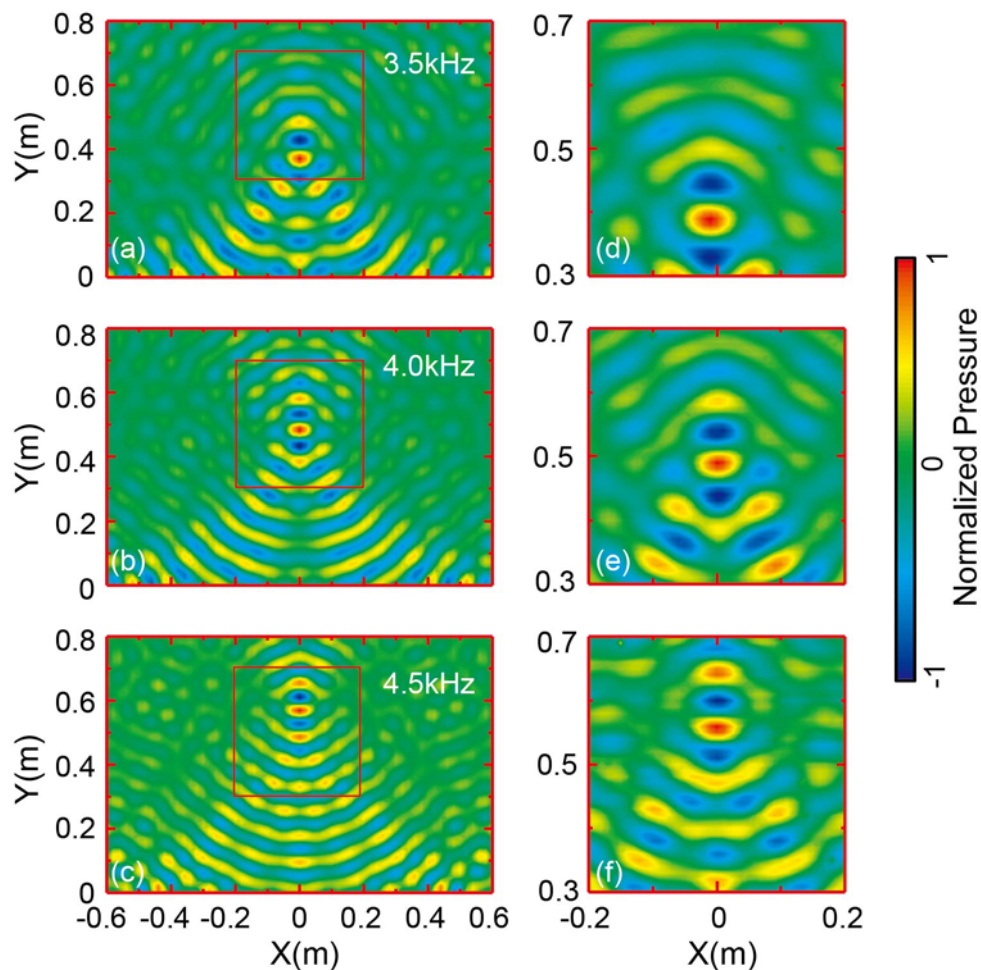
14 and D has been calculated being 0. Furthermore, the good agreement of the theory and
15 simulation result shows the rationality of the derivation.

16 Appendix C. The Pressure and intensity field of sound focusing

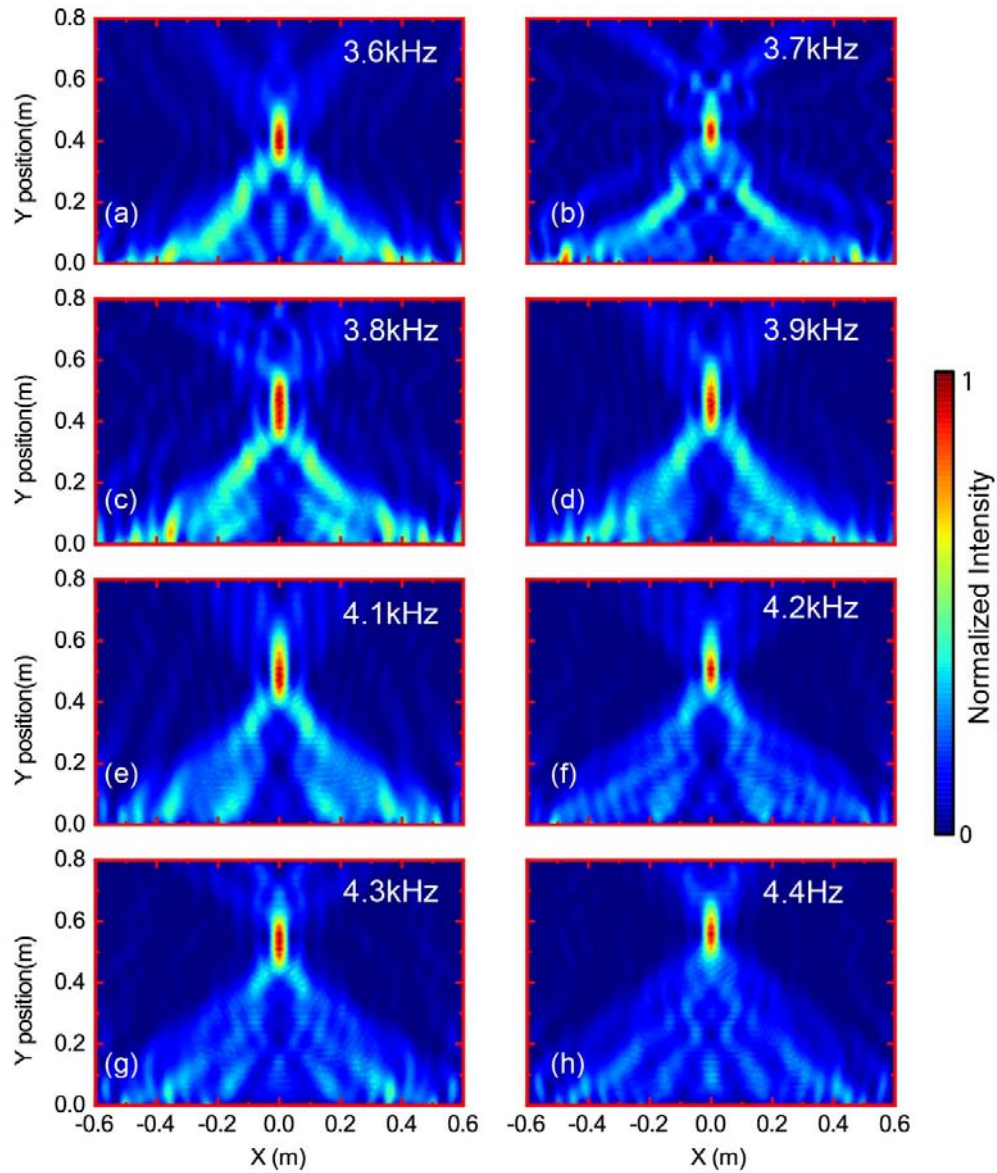
17 To further demonstrate the capability, the pressure pattern of simulation and experiment
18 has been plotted in Fig. C.1 For one thing, it clearly shows the bent wavefronts. For another,
19 along with the frequency increase, the focal spot moves further which is in good agreement
20 with the theoretical analysis.

21 The available frequency band of the designed focal lens ranges from about 3.5 kHz to 4.5

1 kHz, not limited to the three separated frequencies demonstrated in the main text. To
2 intensively verify the flat lens' performance, simulations at other frequencies have been
3 conducted. The simulated acoustic intensity fields in Fig. C.2 shows that the lens has an ideal
4 and uniform performance for sound focusing over a considerable bandwidth.



5
6 Fig. C.1. Comparison of the measured and simulated acoustic pressure fields. (a)-(c) are the
7 simulation results and (d)-(f) are the experimental results. Red boxes in the left sub-figures
8 denote the scanned area.
9

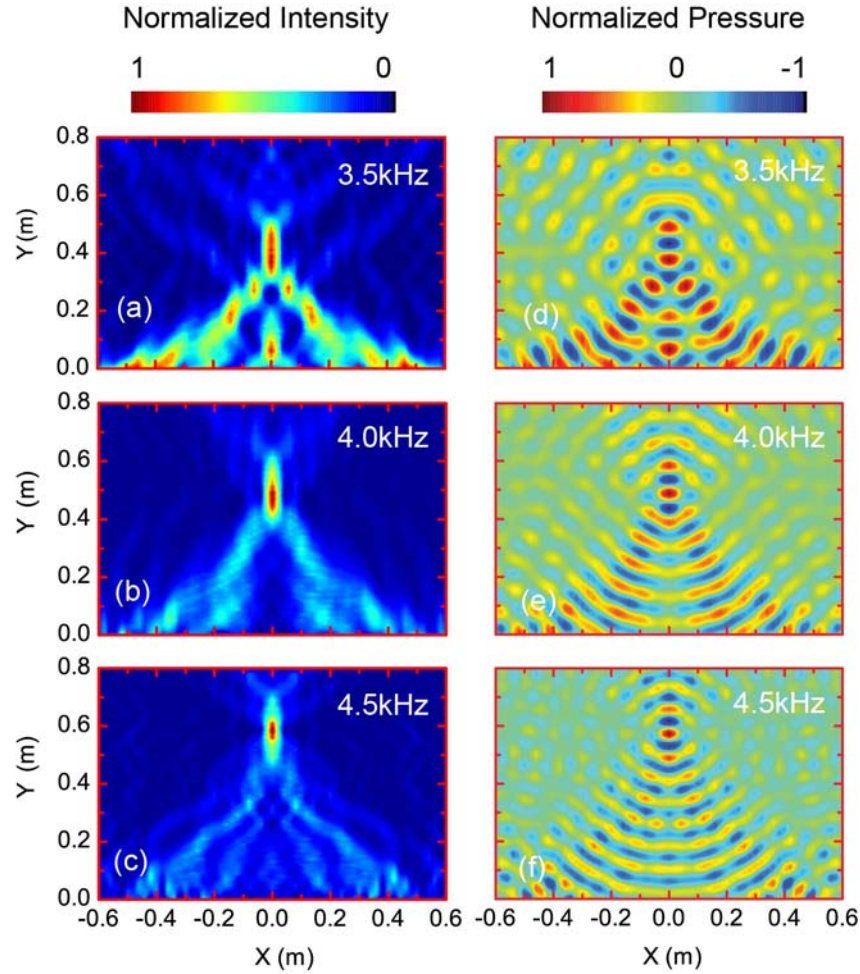


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2 Fig. C.2. The simulated intensity distributions at several different frequencies within 3.5
 3 kHz-4.5 kHz.

4 **Appendix D. Full-wave simulation considering the mechanical properties of the**
 5 **structure**

6 To make sure the design can work in the practical condition, we have conducted a
 7 simulation when considering the mechanical properties of the helical structure. Compared to
 8 the simulation under ideal boundary condition (rigid boundary), the good agreement of the
 9 filed patterns (Fig. D.1) indicates the rationality of the theoretical assumptions.



1
 2 Fig. D.1. Full-wave simulation with real structure properties Fig (a)-(c) are for the intensity
 3 field and figure (d)-(f) are for the pressure field.

4 **Appendix E. Experimental setup.**

5 The unit cells are made from photosensitive resin (material type, Somos GP Plus 14122)
 6 and printed based on laser sintering stereo-lithography. To test the complex transmission
 7 coefficient, a self-made impedance tube is used. The parts of the tube are listed as follows.

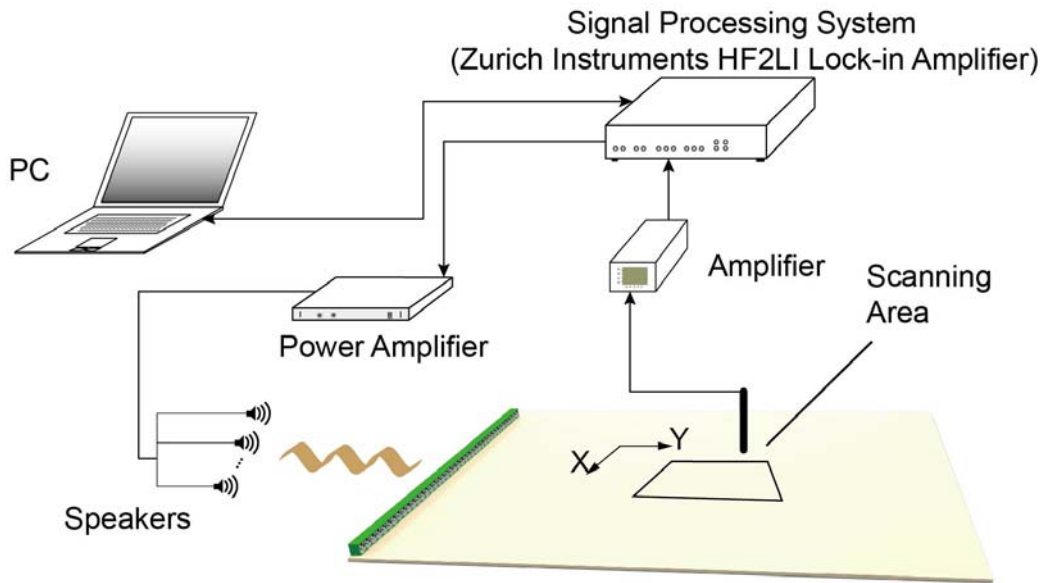
- 8 1. Source, A loudspeaker (Tymphany PMT-40N25AL01-04) driven by a signal
 9 generator (Brüel and Kjær 3560D with 7539 Controller modules and 3109 Output modules)
 10 and a power amplifier (Brüel and Kjær Type 2706)
 11 2. The signal collection, Brüel and Kjær type microphone 1/4 inch (Type 4935)

12 By using the transfer matrix method and the measured complex pressure, the transmission
 13 coefficient and phase delay of the unit cell sample has been calculated.

14 The experiment of broadband sound focusing has been conducted in half space. To
 15 generate an experimental plane wave, thirty loudspeakers with an interval 40 mm are used to
 16 assemble a line source and synchronously driven by Zurich Instruments HF2LI Lock-in
 17 Amplifier and a power amplifier (Brüel and Kjær Type2716C). Forty-one printed unit cells
 18 are fixed by a holey frame and the interval is 32mm. The frame structure is manufactured by

1 the MakerBot Z18 3D printer using PLA material. For the scanning of the field, we used a
2 2-axis mobile platform to carry the microphone (Type 4935) for the point-by-point test. The
3 measurement of the sound pressure is completed by the Lock-in system and software.

4 In the simulation and experiment setup, the lens is positioned along the X-axis; the middle
5 point of the lens is same as the coordinate zero point. To measure the focal spot, the
6 experimentally scanned region is from -0.2 m-0.2 m along the X-axis and 0.3 m-0.7 m along
7 the Y-axis. The plane wave source of speaker array on the other side of X-axis is 70mm away
8 from the lens in a parallel position.

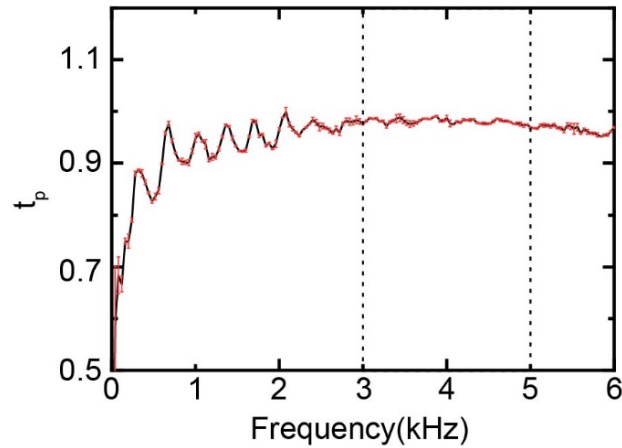


9

10 Fig. E.1. The sound field scanning setup. The Lock-in amplifier controlled by the computer
11 works as both signal generation and data acquisition systems. An array of thirty speakers are
12 in parallel to generate a plane wave positioned 0.07 m away from the metalens. For the
13 scanning of the field, we measure the target region point by point with a step of 0.01 m.

14 **Appendix F. A note for the working frequency range of the impedance tube**

15 We just select 3 kHz-5 kHz in section 3.2, because of the performance of our lab made
16 impedance tube. In low frequency, the tested parameter can be influenced by the thermal and
17 viscous loss of the duct itself; in high frequency, which is higher than the cut frequency, the
18 high-order mode can be generated to make the results lose the precision. When to demonstrate
19 the difference between two cells, the error causes small influence on the outcome. However,
20 when to discuss effective parameters in section 3.2, a more accurate data set is needed. Thus,
21 the range 3 kHz - 5 kHz can be accepted. The performance of the impedance tube can refer to
22 Fig. F.1. We measured its transmission of sound pressure over the spectrum, due to the
23 limited experimental conditions, in the low-frequency range (<3 kHz) it has a large
24 attenuation while a slight fluctuation in the high-frequency range (>5 kHz).



1
2 Fig. F.1. The measured energy transmission of the empty lab-made impedance tube

3
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