# On Realization Theory of Quantum Linear Systems

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## Abstract

The purpose of this paper is to study the realization theory of quantum linear systems. It is shown that for a general quantum linear system its controllability and observability are equivalent and they can be checked by means of a simple matrix rank condition. Based on controllability and observability a specific realization is proposed for general quantum linear systems in which an uncontrollable and unobservable subspace is identified. When restricted to the passive case, it is found that a realization is minimal if and only if it is Hurwitz stable. Computational methods are proposed to find the cardinality of minimal realizations of a quantum linear passive system. It is found that the transfer function G(s) of a quantum linear passive system can be written as a fractional form in terms of a matrix function  $\Sigma(s)$ ; moreover, G(s) is lossless bounded real if and only if  $\Sigma(s)$  is lossless positive real. A type of realization for multi-input-multi-output quantum linear passive systems is derived, which is related to its controllability and observability decomposition. Two realizations, namely the independent-oscillator realization and the chain-mode realization, are proposed for single-input-single-output quantum linear passive systems, and it is shown that under the assumption of minimal realization, the independent-oscillator realization is unique, and these two realizations are related to the lossless positive real matrix function  $\Sigma(s)$ .

Key words: Quantum linear systems, realization theory, controllability, observability.

# 1 Introduction

Linear systems and signals theory has been very useful in the analysis and engineering of dynamical systems. Many fundamental notions have been proposed to characterize dynamical systems from a control-theoretic point of view. For example, controllability describes the ability of steering internal system states by external input, observability refers to the possibility of reconstructing the state-space trajectory of a dynamical system based on its external input-output data. Based on controllability and observability, Kalman canonical decomposition reveals the internal structure of a linear system. This, in particular minimal realization as a very convenient and yet quite natural assumption, is the basis of widely used model reduction methods such as balanced truncation and optimal Hankel norm approximation. Moreover, fundamental dissipation theory has been well established and has been proven very effective in control systems design. All of these have been well documented, see, e.g., Anderson & Vongpanitlerd, 1973; Kailath, 1980; Zhou, Doyle & Glover, 1996.

In recent years there has been a rapid growth in the study of quantum linear systems. Quantum linear systems and signals theory has been proven to be very effective in the study of many quantum systems including quantum optical systems, opto-mechanical systems, cavity quantum electro-magnetic dynamical systems, and atomic ensembles, see, e.g., Wiseman & Milburn, 2010; Zhang, Chen, Bhattacharya, & Meystre, 2010; Massel, et al., 2011; Matyas, et al., 2011. We mention especially that highly-nontrivial quantum passive systems have been proposed as quantum memory devices (Hush, Carvalho, Hedges & James, 2013; Yamamoto & James, 2014), and our results in Section 4 may be useful for designing such

Preprint submitted to Automatica

28 December 2021

<sup>\*</sup> This paper was not presented at any IFAC meeting. This research is supported in part by National Natural Science Foundation of China (NSFC) grant (No. 61374057), a Hong Kong RGC grant (No. 531213) and a grant from the Royal Society of Engineering, UK. Corresponding author Guofeng Zhang. Tel. +852-2766-6936. Fax +852-2764-4382. Email: Guofeng.Zhang@polyu.edu.hk.

memory components. Because of its analytical and computational advantages, the linear setting always serves as an essential starting point for development of a more general theory.

Controllability and observability of quantum linear passive systems have been discussed in Maalouf & Petersen, 2011a; these two properties are used to establish the complex-domain bounded real lemma (Maalouf & Petersen, 2011a, Theorem 6.5) for quantum linear passive systems, which is the basis of quantum  $H^{\infty}$  coherent feedback control of quantum linear passive systems. For a quantum linear passive system it is shown in Guta & Yamamoto, 2013, Lemma 3.1 that controllability is equivalent to observability; moreover, a minimal realization is necessarily Hurwitz stable, Guta & Yamamoto, 2013, Lemma 3.2. In this paper we explore further controllability and observability of quantum linear systems. For general quantum linear systems (not necessarily passive), we show that controllability and observability are equivalent (Proposition 1). Moreover, a simple matrix rank condition is established for checking controllability and observability. Base on this result, a realization of general quantum linear systems is proposed, in which the uncontrollable and unobservable subsystem is identified (Theorem 1). Theorem 1 can be viewed as the complexdomain counterpart of Theorem 3.1 in Yamamoto, 2014 in the real domain. However, it is can be easily seen from the proof of Lemma 1 that the structure of the unitary transformation involved is better revealed in the complex domain. Restricted to the passive case, we show that controllability, observability and Hurwitz stability are equivalent to each other (Lemma 2). Thus, a realization of a quantum linear passive system is minimal if and only if it is Hurwitz stable (Theorem 2). We also derive formulas for calculating the cardinality of minimal realizations of a given quantum linear passive system (Proposition 4) for the single-input-single-output case and Proposition 5 for the multi-input-multi-output case). Finally we show how a given quantum linear system can be written as a fractional form in term of a matrix function  $\Sigma$  (Proposition 3), and for the passive case show that a quantum linear passive system G is lossless bounded real if and only if the corresponding  $\Sigma$  is lossless positive real (Theorem 3). Finally, we also mention that it is possible to perform continuous non-demolition measurements on the output fields, in which case one deals with the problem of quantum filtering (or quantum trajectories as it is sometimes referred to in the physics community); for a discussion on this see e.g., Wiseman & Doherty, 2005, which also includes some early results on stabilizability and detectability of quantum linear systems.

The realization problem of quantum linear systems has been investigated in Nurdin, James & Doherty, 2009, where they showed that a quantum linear system can always be realized by a cascade of one-degree-of-freedom harmonic oscillators with possible direct Hamiltonian couplings among them if necessary. Then in Nurdin,

2010b a necessary and sufficient condition is derived for realizing quantum linear systems via pure cascading only. For the passive case, it is shown in Petersen, 2011 that, under certain conditions on the system matrices, a minimal quantum linear passive system can be realized by a cascade of one-degree-of-freedom harmonic oscillators. These restrictions were removed in Nurdin, 2010b which proves that all quantum linear passive systems can be realised by pure cascading of one-degree-offreedom harmonic oscillators. Model reduction of quantum linear systems has been studied in, e.g., Petersen, 2013, and Nurdin, 2014. In this paper we propose several realizations of quantum linear passive systems. For the multi-input-multi-output (MIMO) case we show that the proposed realization has a close relationship with controllability and observability of the quantum linear passive system (Theorem 4). In the single-inputsingle-output (SISO) case, we propose two realizations, namely the independent-oscillator realization and the chain-mode realization (Theorem 5 and Theorem 6), and finally we show that if the system is Hurwitz stable, these two realizations are related to the lossless positive real  $\Sigma$  mentioned in the previous paragraph (Theorem 7). Finally, it is worthwhile to point out that the issue of realization of quantum linear systems is a bit subtle. According to classical linear systems theory, see e.g., Anderson & Vongpanitlerd, 1973, Zhou, Doyle & Glover, 1996, a state-space model with matrices(A, B, C, D) for a classical linear system can always be implemented physically at least approximately, for example by means of mechanical and electrical devices. So a classical state-space model is always realizable. In the quantum regime, a mathematical model for a quantum linear system may have parameters such as internal Hamiltonian and the coupling L between the system and the external field. However, it might be difficult to physically implement this model directly in terms of these system parameters. Instead, some symplectic or unitary transformation may be applied to the mathematical model so that a new set of system parameters is obtained, on the basis of which the system is physically realized. However, the transformed system is not physically equivalent to the original system but has the same transfer function as the original. That is, the transformed system is a transfer function realization of the original, in same spirit as classical linear realization theory. Therefore in the quantum setting, there can be two distinct quantum linear realization problems, a "strict" problem of realizing the physical operators describing a given linear quantum system (e.g., the Hamiltonian and coupling operators) Nurdin, James & Doherty, 2009, and a "soft" problem of realizing the transfer function of the system Nurdin, 2010b, Petersen, 2011. More discussions can be found in, e.g., Nurdin, James & Doherty, 2009, Nurdin, 2010a, Nurdin, 2010b, Petersen, 2011.

The rest of the paper is organized as follows. Section 2 studies general quantum linear systems; specifically, Subsection 2.1 briefly reviews quantum linear systems,

Subsection 2.2 investigates their controllability and observability, and Subsection 2.3 presents a fractional form for transfer functions of quantum linear systems. Section 3 studies quantum linear passive systems, specifically, Subsection 3.1 introduces quantum linear passive systems, Subsection 3.2 investigates their Hurwitz stability, controllability and observability, Subsection 3.3 studies minimal realizations of quantum linear passive systems, and Subsection 3.4 discusses a relation between the transfer functions G and  $\Sigma$  in the passive case. Section 4 investigates realizations of quantum linear passive systems; specifically, Subsection 4.1 proposes a realization for MIMO quantum linear passive systems, Subsections 4.2.1 and 4.2.2 propose an independent-oscillator realization and a chain-mode realization for SISO quantum linear systems respectively, and Subsection 4.2.3 discusses the uniqueness of the independent-oscillator realization. Section 5 concludes this paper.

Notation. m is the number of input channels, and n is the number of degrees of freedom of a given quantum linear system, namely, the number of system oscillators. Given a column vector of complex numbers or operators  $x = [x_1 \cdots x_k]^T$ , define  $x^{\#} = [x_1^* \cdots x_k^*]^T$ , where the asterisk \* indicates complex conjugation or Hilbert space adjoint. Denote  $x^{\dagger} = (x^{\#})^T$ . Furthermore, define a column vector  $\breve{x}$  to be  $\breve{x} = [x^T \ (x^{\#})^T]^T$ . Let  $I_k$  be an identity matrix and  $0_k$  a zero square matrix, both of dimension k. Define  $J_k = \text{diag}(I_k, -I_k)$ . Then for a matrix  $X \in \mathbb{C}^{2j \times 2k}$ , define  $X^{\flat} = J_k X^{\dagger} J_j$ . Given two constant matrices  $U, V \in \mathbb{C}^{r \times k}$ , define  $\Delta(U, V) = [U V; V^{\#} U^{\#}]$ . Given two operators A and B, their commutator is defined to be [A, B] = AB - BA. " $\iff$ " means if and only if. Finally,  $\operatorname{Spec}(X)$  denotes the set of all *distinct* eigenvalues of the matrix X,  $\sigma(X)$  denotes the diagonal matrix with diagonal entries being the non-zero singular values of the matrix X,  $\operatorname{Ker}(X)$  denotes the null space of the matrix X, and Range (X) denotes the space spanned by the columns of the matrix X.

## 2 Quantum linear systems

In this section, we first introduce quantum linear systems in Subsection 2.1, after that we discuss their controllability and observability in Subsection 2.2, and finally we study their transfer functions in Subsection 2.3.

#### 2.1 Quantum linear systems

In this subsection quantum linear systems are briefly described in terms of the (S, L, H) language, Gough & James, 2009. More discussions on quantum linear systems can be found in, e.g., Wiseman & Milburn, 2010; Doherty & Jacobs, 1999; Zhang & James, 2012; Tezak, et al., 2012.

The open quantum linear system G studied in this paper consists of n interacting quantum harmonic oscillators driven by m input boson fields. Each oscillator j has an annihilation operator  $a_j$  and a creation operator  $a_i^*$ ;  $a_j$ and  $a_i^*$  are operators on the system space  $\mathfrak{h}$  which is an infinite-dimensional Hilbert space. The operators  $a_j, a_k^*$ satisfy the canonical commutation relations:  $[a_j, a_k^*] =$  $\delta_{jk}$ . Denote  $\mathbf{a} \equiv [a_1 \cdots a_n]^T$ . Then the initial (that is, before the interaction between the system and the input boson fields) system Hamiltonian H can be written as H =  $(1/2)\mathbf{\check{a}}^{\dagger}\Omega\mathbf{\check{a}}$ , where  $\mathbf{\check{a}} = [\mathbf{a}^T \ (\mathbf{a}^{\#})^T]^T$  as introduced in the Notation part, and  $\Omega = \Delta(\Omega_-, \Omega_+) \in \mathbb{C}^{2n \times 2n}$ is a Hermitian matrix with  $\Omega_-, \Omega_+ \in \mathbb{C}^{n \times n}$ . L in the (S, L, H) language describes the coupling of the system harmonic oscillators to the input boson fields. The coupling is *linear* and can be written as  $L = [C_- C_+]\mathbf{\breve{a}}$  with  $C_-, C_+ \in \mathbb{C}^{m \times n}$ . Finally, in the linear setting S in the (S, L, H) language is taken to be a constant *unitary* matrix in  $\check{\mathbb{C}}^{m \times m}$ .

Each input boson field j has an annihilation operator  $b_j(t)$  and a creation operator  $b_j^*(t)$ , which are operators on an infinite-dimensional Hilbert space  $\mathfrak{F}$ . Let  $\mathbf{b}(t) \equiv [b_1(t) \cdots b_m(t)]^T$ . The operators  $b_j(t)$  and their adjoint operators  $b_j^*(t)$  satisfy the commutation relations  $[b_j(t), b_k^*(r)] = \delta_{jk}\delta(t-r)$  for all  $j, k = 1, \ldots, m, \forall t, r \in \mathbb{R}$ . For each  $j = 1, \ldots, m$ , the j-th input field can also be represented in the integral form  $B_j(t) \equiv \int_0^t b_j(r)dr$ , whose Ito increment is  $dB_j(t) \equiv B_j(t+dt) - B_j(t)$ . Denote  $B(t) \equiv [B_1(t) \cdots B_m(t)]^T$ . The gauge process can be defined by  $\Lambda_{jk}(t) = \int_0^t b_j^*(r)b_k(r)dr$ ,  $(j, k = 1, \ldots, m)$ . The field studied in this paper is assumed to be canonical, that is, the field operators  $B_j(t), B_k^*(t), \Lambda_{rl}(t)$ satisfy the following Ito table:

×	$dB_k$	$d\Lambda_{kl}$	$dB_l^*$	dt
$dB_i$	0	$\delta_{ik} dB_l$	$\delta_{il}dt$	0
$d\Lambda_{ij}$	0	$\delta_{jk} d\Lambda_{il}$	$\delta_{jl} dB_i^*$	0
$dB_j^*$	0	0	0	0
dt	0	0	0	0

Under mild assumptions, the temporal evolution of the open quantum linear system G can be described in by means of the following quantum stochastic differential equation (QSDE):

$$dU(t) = \left\{ -\left(L^{\dagger}L/2 + iH\right)dt + dB^{\dagger}(t)L - L^{\dagger}SdB(t) + \operatorname{Tr}[(S - I)d\Lambda^{T}(t)] \right\} U(t), \quad t > 0,$$

with U(0) = I being the identity operator. Let X be an operator on the system space  $\mathfrak{h}$ . Then the temporal evolution of X, denoted  $X(t) \equiv U(t)^*(X \otimes I)U(t)$ , is governed by the following QSDE:

$$dX(t) = \mathcal{L}_{L,H}(X(t))dt + dB^{\dagger}(t)S^{\dagger}[X(t), L(t)]$$
(1)  
+ [L^{\dagger}(t), X(t)]SdB(t) + Tr[(S^{\dagger}X(t)S - X(t))d\Lambda^{T}(t)],

where the Lindblad operator  $\mathcal{L}_{L,H}(X(t))$  is

$$\begin{split} \mathcal{L}_{L,H}(X(t)) &\equiv -i[X(t), H(t)] + \frac{1}{2}L^{\dagger}(t)[X(t), L(t)] \\ &+ \frac{1}{2}[L^{\dagger}(t), X(t)]L(t). \end{split}$$

Note that X(t) is an operator on the joint space  $\mathfrak{h} \otimes \mathfrak{F}$ .

Let  $b_{\text{out},j}(t)$  denote the *j*-th field after interacting with the system, and  $B_{\text{out},j}(t) \equiv \int_0^t b_{\text{out},j}(r)dr$ . We have  $B_{\text{out},j}(t) = U^*(t) (I \otimes B_j(t)) U(t)$ . Denote  $B_{\text{out}}(t) \equiv [B_{\text{out},1}(t), \cdots B_{\text{out},m}(t)]^T$ . Then in compact form the output field equation is  $dB_{\text{out}}(t) = L(t)dt + SdB(t)$ . Substituting  $H = (1/2)\breve{a}^{\dagger}\Omega\breve{a}$  and  $L = [C_- C_+]\breve{a}$  into (1) we have a quantum *linear* system:

$$d\mathbf{\breve{a}}(t) = \mathcal{A}\mathbf{\breve{a}}(t)dt + \mathcal{B}d\mathbf{\breve{B}}(t), \tag{2}$$

$$dB_{\rm out}(t) = \mathcal{C}\breve{\mathbf{a}}(t)dt + \mathcal{D}dB(t), \qquad (3)$$

in which

$$\mathcal{A} = -\frac{1}{2}C^{\flat}C - iJ_n\Omega, \ \mathcal{B} = -C^{\flat}\Delta(S, 0_{m \times m}),$$
  
$$\mathcal{C} = \Delta(C_-, \ C_+) \equiv C, \ \mathcal{D} = \Delta(S, 0_{m \times m}).$$
(4)

Clearly, the quantum linear system (2)-(3) is parameterized by constant matrices  $S, C, \Omega$ . In the sequel, we use the notation  $G \sim (S, C, \Omega)$  for the quantum linear system (2)-(3) with parameters given in (4).

The constant matrices  $\mathcal{A}, \mathcal{B}, \mathcal{C}, \mathcal{D}$  in (4) satisfy the fundamental relations  $\mathcal{A} + \mathcal{A}^{\flat} + \mathcal{C}^{\flat}\mathcal{C} = 0$ ,  $\mathcal{B} = -\mathcal{C}^{\flat}\mathcal{D}, \mathcal{D}^{\flat}\mathcal{D} = I_{2m}$ . These equations are often called *physical realizability* conditions of quantum linear systems. More discussions on physical realizability of quantum linear systems can be found in, e.g., James, Nurdin & Petersen, 2008; Zhang & James, 2011; Zhang & James, 2012.

#### 2.2 Controllability and observability

In this subsection we study controllability and observability of quantum linear systems above introduced.

Let X be an operator on the system space  $\mathfrak{h}$ . Denote by  $\langle X(t) \rangle$  the expected value of X(t) with respect to the initial joint system-field state (which is a unit vector in the Hilbert space  $\mathfrak{h} \otimes \mathfrak{F}$ ). Then (2)-(3) gives rise to the

following *classical* linear system

$$\frac{d\langle \mathbf{\breve{a}}(t) \rangle}{dt} = \mathcal{A} \langle \mathbf{\breve{a}}(t) \rangle + \mathcal{B} \langle \mathbf{\breve{b}}(t) \rangle, \tag{5}$$

$$\frac{d\langle \breve{\mathbf{b}}_{\text{out}}(t) \rangle}{dt} = \mathcal{C} \left\langle \breve{\mathbf{a}}(t) \right\rangle + \mathcal{D} \left\langle \breve{\mathbf{b}}(t) \right\rangle.$$
(6)

**Definition 1** The quantum linear system  $G \sim (S, C, \Omega)$ is said to be Hurwitz stable (resp. controllable, observable) if the corresponding classical linear system (5)-(6) is Hurwitz stable (resp. controllable, observable).

Due to the special structure of quantum linear systems, we have the following result concerning their controllability and observability.

**Proposition 1** For the quantum linear system  $G \sim (S, C, \Omega)$ , the following statements are equivalent:

(i) G is controllable;
(ii) G is observable;
(iii) rank(O<sub>s</sub>) = 2n, where

$$\mathbf{O}_{s} \equiv \begin{bmatrix} C \\ CJ_{n}\Omega \\ \vdots \\ C\left(J_{n}\Omega\right)^{2n-1} \end{bmatrix}.$$
 (7)

**Proof.** Firstly, we show that (i) and (ii) are equivalent by a contradiction argument, that is, we establish that uncontrollability is equivalent to unobservability. Assume G is not observable. In what follows we show that G is not controllable. Indeed, if G is not observable, then by the classical control theory (see. e.g., Zhou, Doyle & Glover, 1996, Theorems 3.3) there exist a scalar  $\lambda$  and a non-zero vector  $v \in \mathbb{C}^{2n}$  such that  $Av = \lambda v$  and  $\mathcal{C}v = 0$ . So  $J_n\Omega v = i\lambda v$  and Cv = 0. Let  $u = J_n v$  and  $\mu = -\lambda^*$ . Then  $u^{\dagger}\mathcal{B} = -v^{\dagger}C^{\dagger}J_m = 0$ , and  $u^{\dagger}\mathcal{A} = -(J_nv)^{\dagger}(C^{\flat}C/2 + iJ_n\Omega) = -(J_nv)^{\dagger}iJ_n\Omega = \mu u^{\dagger}$ . By a standard result in classical control theory, (see. e.g., Zhou, Doyle & Glover, 1996, Theorems 3.1), G is not controllable. Similarly, if G is not controllable, then it can be shown that it is not observable.

Secondly, we show that (ii) and (iii) are equivalent. Assume G is observable. In the following we show that for an arbitrary  $v \in \mathbb{C}^{2n}$  such that  $\mathbf{O}_s v = 0$ , we have v = 0, thus establishing (iii). Indeed, if  $\mathbf{O}_s v = 0$ , we have  $\mathcal{C}v = \mathcal{C}v = 0$  and  $\mathcal{C}(J_n\Omega)^k v = 0$ ,  $k = 1, \ldots, 2n-1$ . Moreover,

$$\begin{aligned} \mathcal{C}\mathcal{A}v &= -C\left(C^{\flat}C/2 + iJ_n\Omega\right)v = -iCJ_n\Omega v = 0,\\ \mathcal{C}\mathcal{A}^2v &= -C\left(C^{\flat}C/2 + iJ_n\Omega\right)^2 v = C(J_n\Omega)^2 v = 0,\\ &\vdots\\ \mathcal{C}\mathcal{A}^{2n-1}v &= C(J_n\Omega)^{2n-1}v = 0. \end{aligned}$$

Because G is observable, v = 0. (iii) is established. In a similar way, it can be shown that, if rank $(\mathbf{O}_s) = 2n$ , then G is observable.  $\Box$ 

Proposition 1 tells us that the controllability and observability of the quantum linear system  $G \sim (S, C, \Omega)$  are equivalent; moreover they can be determined by checking the rank of the matrix  $\mathbf{O}_s$ .

On the basis of Proposition 1, we have the following result about the uncontrollable and unobservable subspace of the quantum linear system  $G \sim (S, C, \Omega)$ .

**Proposition 2** Let  $\mathbf{C} \equiv [\mathcal{B} \ \mathcal{A}\mathcal{B} \ \cdots \ \mathcal{A}^{2n-1}\mathcal{B}]$  and  $\mathbf{O} \equiv [\mathcal{C}^T \ (\mathcal{C}\mathcal{A})^T \ \cdots \ (\mathcal{C}\mathcal{A}^{2n-1})^T]^T$  be the controllability and observability matrices of the quantum linear system  $G \sim (S, C, \Omega)$  respectively. Then (in the terminology of modern control theory, Anderson & Vongpanitlerd, 1973; Kailath, 1980; Zhou, Doyle & Glover, 1996) the following statements hold:

- (i) The unobservable subspace is  $\text{Ker}(\mathbf{O}) = \text{Ker}(\mathbf{O}_s)$ , where Ker(X) denotes the null space of the matrix X, as introduced in the Notation part.
- (ii) The uncontrollable subspace is  $\operatorname{Ker}(\mathbf{C}^{\dagger}) = \operatorname{Ker}(\mathbf{O}_{\mathbf{s}}J_{n}).$
- (iii) The uncontrollable and unobservable subspace is  $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n).$

Proposition 2 can be established in the similar way as Proposition 1.

Propositions 1 and 2 appear purely algebraic. Nevertheless, they have interesting and important physical consequences. We begin with the following lemma.

**Lemma 1** The dimension of the space Ker  $(\mathbf{O}_s) \cap \text{Ker} (\mathbf{O}_s J_n)$ is even. Let the dimension of Ker  $(\mathbf{O}_s) \cap \text{Ker} (\mathbf{O}_s \mathbf{J}_n)$  be 2l for some nonnegative integer l. There exists a matrix  $V = [V_1 V_2]$  with  $V_1 \in \mathbb{C}^{2n \times 2l}$  and  $V_2 \in \mathbb{C}^{2n \times 2(n-l)}$ such that

 $\operatorname{Range}(V_1) = \operatorname{Ker}\left(\mathbf{O}_s\right) \cap \operatorname{Ker}\left(\mathbf{O}_s J_n\right),\tag{8}$ 

$$VV' = V'V = I_{2n},\tag{9}$$

$$V^{\dagger}J_{n}V = \begin{bmatrix} J_{l} & 0\\ 0 & J_{n-l} \end{bmatrix}.$$
 (10)

The proof is given in the Appendix.

We are ready to state the main result of this section.

**Theorem 1** Let V be the matrix defined in Lemma 1. If Range(V<sub>1</sub>) is an invariant space under the linear transformation of  $\Omega$ , then the transformed system  $\begin{bmatrix} \breve{a}_{DF} \\ \breve{a}_{D} \end{bmatrix} \equiv$ 

 $V^{\dagger}\breve{a}$  has the following dynamics:

$$d\check{a}_{DF}(t) = -iJ_l V_1^{\dagger} \Omega V_1 \check{a}_{DF}(t) dt, \qquad (11)$$
  
$$d\check{a}_D(t) = -\left( (CV_2)^{\flat} (CV_2)/2 + iJ_{n-l} V_2^{\dagger} \Omega V_2 \right) \check{a}_D(t) dt$$

$$(CV_2)^{\flat} \mathcal{D}d\check{B}(t),\tag{12}$$

$$d\check{B}_{\rm out}(t) = (CV_2)\check{a}_D(t)dt + \mathcal{D}d\check{B}(t).$$
(13)

**Proof.** Because  $\operatorname{Range}(V_1) = \operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$ , the coupling operator of the transformed mode  $[\check{a}_{DF}^T \check{a}_D^T]^T$  is  $\mathcal{C}V = [0 \ CV_2]$ . Moreover, because  $\operatorname{Range}(V_1)$  is an invariant space under the linear transformation of  $\Omega$ , there exists a matrix Y such that  $\Omega V_1 = V_1 Y$ . We have  $V_1^{\dagger} \Omega V_2 = Y^{\dagger} V_1^{\dagger} V_2 = 0$  where (9) is used. This, together with (10), gives

$$V^{\dagger}J_{n}\Omega V = V^{\dagger}J_{n}VV^{\dagger}\Omega V = \begin{bmatrix} J_{l}V_{1}^{\dagger}\Omega V_{1} & 0\\ 0 & J_{n-l}V_{2}^{\dagger}\Omega V_{2} \end{bmatrix}.$$

That is, the transformed system with mode  $[\breve{a}_{DF}^T \ \breve{a}_D^T]^T$  has the dynamics (11)-(13).  $\Box$ 

Remark 1. By (11), the modes  $\check{a}_{DF}$  evolve unitarily as an isolated system. In literature such isolated modes embedded in an open quantum system is often called *decoherence-free* modes, see, e.g., Ticozzi & Viola, 2008, Ticozzi & Viola, 2009, Yamamoto, 2014. Theorem 1 can be viewed as the complex-domain counterpart of Theorem 3.1 in Yamamoto, 2014 in the real domain. However, with the help of the matrix  $\mathbf{O}_s$ , matters are simplified; moreover, it can be seen from the proof of Lemma 1 in the Appendix that the structure of the linear transformation matrix V is better revealed with the help of  $\mathbf{O}_s$ and in the complex domain.

Finally, from the proof of Lemma 1 it can be seen that the dimension of the space Ker(C) is also even. Moreover we have the following corollary which shows that under some conditions the unobservable and uncontrollable subspace is exactly Ker(C).

**Corollary 1** Let the dimension of the space  $\operatorname{Ker}(C)$  be 2r. Let a matrix  $T \in \mathbb{C}^{2n \times 2r}$  be such that  $\operatorname{Range}(T) = \operatorname{Ker}(C)$ . If  $J_nT = TJ_r$  and  $\operatorname{Range}(T)$  is an invariant space under the linear transformation of  $\Omega$ , then  $\operatorname{Ker}(C) = \operatorname{Ker}(\mathbf{O_s}) \cap \operatorname{Ker}(\mathbf{O_s}J_n)$ . **Proof.** Clearly,  $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n) \subset \operatorname{Ker}(\mathbf{O}_s) \subset$   $\operatorname{Ker}(C)$ . Thus it is sufficient to show that  $\operatorname{Ker}(C) \subset$   $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$ . However  $\operatorname{Range}(T) = \operatorname{Ker}(C)$ , therefore we show that  $\operatorname{Range}(T) \subset \operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$ . Because  $\operatorname{Range}(T)$  is invariant with respect to a linear transformation  $\Omega$ , there exist matrix Y such that  $\Omega T = TY$ . This, together with  $J_n T = TJ_r$ , gives  $C(J_n\Omega)T = CTJ_rY = 0$ . Similarly, for all  $k \geq 1$ ,  $C(J_n\Omega)^k T = 0$ . That is,  $\mathbf{O}_s T = 0$ . Moreover,  $\mathbf{O}_s J_n T = \mathbf{O}_s TJ_r = 0$ . Consequently  $\operatorname{Ker}(C) =$   $\operatorname{Range}(T) \subset \operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$ . This together with  $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n) \subset \operatorname{Ker}(C)$  yields  $\operatorname{Ker}(C) =$  $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$ .  $\Box$ 

Corollary 1 can be regarded as the complex-domain counterpart of Proposition 3.1 in Yamamoto, 2014 in the real domain.

### 2.3 Transfer functions

In this subsection the concept of transfer functions is introduced and some properties of transfer functions for quantum linear systems are presented.

In classical linear systems theory, a transfer function H(s) is a function which specifies the input-output relation in the frequency domain. Given a transfer function H(s), if there exist matrices A, B, C, D such that  $H(s) = D + C(sI - A)^{-1}B$ , then we say the transfer function H(s) can be realized by a state-space model

$$dx(t) = Ax(t)dt + Bdu(t), \tag{14}$$

$$dy(t) = Cx(t)dt + Ddu(t).$$
(15)

Transfer functions for quantum linear systems have been defined in a similar way, see e.g., Yanagisawa & Kimura, 2003, Gough, James & Nurdin, 2010, Shaiju & Petersen, 2012. Clearly, the linear dynamical system (2)-(3) with system matrices  $\mathcal{A}, \mathcal{B}, \mathcal{C}, \mathcal{D}$  in (4) is a realization of the transfer function

$$G(s) \equiv \mathcal{D} + \mathcal{C}(sI - \mathcal{A})^{-1}\mathcal{B}.$$
 (16)

It should be noticed that, although the dynamics (14)-(15) and (2)-(3) look formally similar, they are essentially different in nature. The linear dynamics (14)-(15) describe a classical system where x(t), u(t), y(t) are time-domain functions. In contrast, the linear dynamics (2)-(3) describe a quantum system where  $\breve{a}(t), \breve{B}(t), \breve{B}_{out}(t)$  are operators on Hilbert spaces, cf. Subsection 2.1.

As in the classical setting, a transfer function may have many different forms of realisations, we introduce the following concept. **Definition 2** Two realizations are said to be equivalent if there determine the same transfer function.

Next we study properties of the transfer function G(s) defined in (16). G(s) has the following fundamental property, see, e.g., Zhang & James, 2013, Eq. (24):

$$G(i\omega)^{\flat}G(i\omega) = G(i\omega)G(i\omega)^{\flat} = I_{2m}, \quad \forall \omega \in \mathbb{R}.$$
(17)

Interestingly, the transfer function G(s) defined in (16) can be written into a fractional form.

**Proposition 3** The transfer function G(s) determined by a Hurwitz stable quantum linear system  $G \sim (S, C, \Omega)$ can be written in the following fractional form

$$G(s) = (I - \Sigma(s))(I + \Sigma(s))^{-1}\Delta(S, 0), \quad (18)$$

where

$$\Sigma(s) \equiv \frac{1}{2}C(sI + iJ_n\Omega)^{-1}C^{\flat}, \quad \forall \operatorname{Re}[s] > 0.$$
 (19)

**Proof.** Because the system  $G \sim (S, C, \Omega)$  is Hurwitz stable, all the eigenvalues of the matrix A have strictly negative real part, therefore the matrix sI - A is invertible for all  $\operatorname{Re}[s] > 0$ . Moreover, for all  $\operatorname{Re}[s] > 0$ , by the Woodbury matrix inversion formula,

$$(sI - \mathcal{A})^{-1}$$
  
=  $(sI + iJ_n\Omega)^{-1} - \frac{1}{2} (sI + iJ_n\Omega)^{-1} C^{\flat}$   
 $\times \left(I + \frac{1}{2}C (sI + iJ_n\Omega)^{-1} C^{\flat}\right)^{-1} C (sI + iJ_n\Omega)^{-1}.$ 

As a result, for all  $\operatorname{Re}[s] > 0$ ,  $I - C(sI - \mathcal{A})^{-1}C^{\flat} = (I - \Sigma(s))(I + \Sigma(s))^{-1}$  with  $\Sigma(s)$  as defined in (19). Consequently,  $G(s) = (I - \Sigma(s))(I + \Sigma(s))^{-1}\Delta(S, 0)$ .  $\Box$ 

#### 3 Quantum linear passive systems

In this section quantum linear passive systems are studied. This type of systems is introduced in Subsection 3.1. Stability, controllability and observability are investigated in Subsection 3.2, while minimal realizations of quantum linear passive systems are studied in Subsection 3.3. The relation between G(s) and  $\Sigma(s)$  in the passive setting is discussed in Subsection 3.4.

#### 3.1 Quantum linear passive systems

If the matrices  $C_+ = 0$  and  $\Omega_+ = 0$ , the resulting system, parameterized by matrices S,  $C_-$ ,  $\Omega_-$ , is often said to be a quantum linear *passive* system. In this case, it

can be described entirely in terms of annihilation operators. Actually a quantum linear passive system has the following form:

$$d\mathbf{a}(t) = A\mathbf{a}(t)dt - C_{-}^{\dagger}SdB(t), \qquad (20)$$

$$dB_{out}(t) = C_{-}\mathbf{a}(t)dt + SdB(t).$$
<sup>(21)</sup>

in which  $A \equiv -\frac{1}{2}C_{-}^{\dagger}C_{-} - i\Omega_{-}$ . Clearly, the transfer function determined from  $G \sim (S, C_{-}, \Omega_{-})$  is

$$G(s) \equiv S - C_{-}(sI - A)^{-1}C_{-}^{\dagger}S.$$
 (22)

Define

$$\Sigma(s) \equiv \frac{1}{2} C_{-} (sI + i\Omega_{-})^{-1} C_{-}^{\dagger}.$$
 (23)

Then, in analog to Proposition 3, we have

$$G(s) = (I - \Sigma(s))(I + \Sigma(s))^{-1}S.$$
 (24)

In the passive case, Eq. (17) reduces to

$$G(i\omega)^{\dagger}G(i\omega) = G(i\omega)G(i\omega)^{\dagger} = I_m, \quad \forall \omega \in \mathbb{R}.$$
 (25)

#### 3.2 Stability, controllability, and observability

In this subsection we study stability of quantum linear passive systems. In particular, we show that a quantum linear passive system  $G \sim (S, C_{-}, \Omega_{-})$  is Hurwitz stable if and only if it is observable and controllable.

**Lemma 2** The following statements for a quantum linear passive system  $G \sim (S, C_{-}, \Omega_{-})$  are equivalent:

(i) G is Hurwitz stable;
(ii) G is observable;
(iii) G is controllable.

**Proof.** (i)  $\rightarrow$  (ii). Clearly,  $X = I_n > 0$  is the unique solution to the following Lyapunov equation

$$A^{\dagger}X + XA + C_{-}^{\dagger}C_{-} = 0.$$
 (26)

According to Zhou, Doyle & Glover, 1996, Lemma 3.18,  $(C_{-}^{\dagger}C_{-}, A)$  is observable, so  $(C_{-}, A)$  is observable. That is, G is observable.

(ii)  $\rightarrow$  (i). Because  $X = I_n > 0$  is a solution to Eq. (26),  $C_{-}^{\dagger}C_{-} \geq 0$  and  $(C_{-}^{\dagger}C_{-}, A)$  is observable, by Zhou, Doyle & Glover, 1996, Lemma 3.19, A is Hurwitz stable.

The equivalence between (ii) and (iii) has been established in Proposition 1.  $\Box$ 

Remark 2. An alternative proof of the equivalence between (ii) and (iii) is given in Guta & Yamamoto, 2013, Lemma 3.1. An alternative proof of (ii)  $\rightarrow$  (i) is given in Guta & Yamamoto, 2013, Lemma 3.2.

#### 3.3 Minimal realization

In this subsection we study minimal realization of a given quantum linear passive system. We first introduce the concept of minimal realization.

**Definition 3** Given a transfer function, let  $G \sim (S, C_-, \Omega_-)$ , or equivalently (20)-(21), be a quantum linear passive system which realizes the transfer function. If  $G \sim (S, C_-, \Omega_-)$  is both controllable and observable, then it is said to be a minimal realization for the transfer function.

The following result is an immediate consequence of Lemma 2.

**Theorem 2** (20)-(21) is a minimal realization of the transfer function G(s) defined in Eq. (22) if and only if it is Hurwitz stable.

In what follows we study the following problem concerning minimal realization.

**Problem 1** Given a quantum linear passive system  $G \sim (S, C_-, \Omega_-)$  which may not be Hurwitz stable, it may have a subsystem  $(S, C_{\min}, \Omega_{\min})$  which is Hurwitz stable. In this case, let  $n_{\min}$  be the number of system oscillators in the minimal realization of  $(S, C_{\min}, \Omega_{\min})$ . How to compute  $n_{\min}$ ?

### 3.3.1 The single-input-single-output (SISO) case

Given a SISO quantum linear passive system  $G \sim (S, C_-, \Omega_-)$ , let the spectral decomposition of  $\Omega_-$  be  $\Omega_- = \sum_{\omega \in \operatorname{spec}(\Omega_-)} \omega P_{\omega}$ , where  $P_{\omega}$  denotes the projection onto the eigenspace of the eigenvalue  $\omega$  of  $\Omega_-$ . Define

$$\sigma(\Omega_{-}, C_{-}) \equiv \{ \omega \in \operatorname{spec}(\Omega_{-}) : C_{-} P_{\omega} C_{-}^{\dagger} \neq 0 \}.$$
 (27)

The following result shows that the size of the set  $\sigma(\Omega_-, C_-)$  is nothing but  $n_{\min}$ .

**Proposition 4** Given a SISO quantum linear passive system  $G \sim (S, C_{-}, \Omega_{-})$ , the number  $n_{min}$  of oscillators of a minimal realization  $(S, C_{min}, \Omega_{min})$  is equal to the size of the set  $\sigma(\Omega_{-}, C_{-})$  defined in (27).

The proof is given in the Appendix.

3.3.2 The multi-input-multi-output (MIMO) case

The following result is the MIMO version of Proposition 4.

**Proposition 5** For a MIMO quantum linear passive system  $G \sim (S, C_{-}, \Omega_{-})$ , let the distinctive eigenvalues of  $\Omega_{-}$  be  $\omega_1, \ldots, \omega_r$ , each with algebraic multiplicity  $\tau_i$  respectively,  $i = 1, \ldots, r$ . Define  $\Lambda_i = \omega_i I_{\tau_i}, i = 1, \ldots, r$ . Assume

$$\Omega_{-} = \begin{bmatrix} \Lambda_1 & 0 \\ & \ddots \\ 0 & \Lambda_r \end{bmatrix}.$$

Accordingly partition  $C_{-} = [C_1 \quad C_2 \quad \cdots \quad C_r]$  with  $C_i$  having  $\tau_i$  columns,  $i = 1, \ldots, r$ . Then  $n_{\min} = \sum_{i=1}^r \text{column rank}[C_i]$ . In particular, if  $\tau_i = 1$  for all  $i = 1, \ldots, r$ , that is, all poles of  $\Omega_{-}$  are simple poles, then  $n_{\min} = \{\omega_i \in \text{spec}(\Omega_{-}) : \text{Tr}[C_-P_{\omega_i}C_{-}^{\dagger}] \neq 0\}$ , as given in Proposition 4.

The construction in Proposition 5 is essentially the Gilbert's realization. Its proof follows the discussions in Kailath, 1980, Sec. 6.1 or Zhou, Doyle & Glover, 1996, Sec. 3.7. The details are omitted.

3.4 G(s) and  $\Sigma(s)$ 

In this subsection we explore a further relation between G(s) and  $\Sigma(s)$  defined in Eqs. (22) and (23) respectively.

We first review the notions of lossless bounded real and lossless positive real. The bounded real lemma for quantum linear passive systems has been established in Maalouf & Petersen, 2011a. Dissipation theory for more general quantum linear systems has been studied in James, Nurdin & Petersen, 2008, Zhang & James, 2011, while the nonlinear case has been studied in James & Gough, 2010.

**Definition 4** (Lossless Bounded Real, Maalouf & Petersen, 2011a, Definition 6.3.) A quantum linear passive system  $G = (S, C_{-}, \Omega_{-})$  is said to be lossless bounded real if it is Hurwitz stable and Eq. (25) holds.

According to Definition 4, a Hurwitz stable quantum linear passive system is naturally lossless bounded real, as derived in Maalouf & Petersen, 2011a.

Positive real functions have been studied extensively in classical (namely, non-quantum) control theory, see, e.g., Anderson & Vongpanitlerd, 1973. Here we state a complex-domain version of positive real functions.

**Definition 5** (Lossless Positive Real.) A function  $\Xi(s)$ is said to be positive real if it is analytic and satisfies  $\Xi(s) + \Xi(s)^{\dagger} \ge 0$ ,  $\forall \operatorname{Re}[s] > 0$ . Moreover,  $\Xi(s)$  is called lossless positive real if is positive real and satisfies  $\Xi(i\omega) + \Xi(i\omega)^{\dagger} = 0$ , where  $i\omega$  is not a pole of  $\Xi(s)$ .

The following result relates the lossless bounded realness of a quantum linear passive system  $G \sim (S, C_{-}, \Omega_{-})$  to the lossless positive realness of  $\Sigma(s)$  defined in Eq. (23). **Theorem 3** If a quantum linear passive system  $G \sim (S, C_{-}, \Omega_{-})$  is minimal, then

(i) G(s) is lossless bounded real.
(ii) Σ(s) defined in Eq. (23) is lossless positive real.

In fact, properties (i) and (ii) are equivalent.

**Proof.** (i). Without loss of generality, assume  $S = I_m$ . Because  $G \sim (I, C_-, \Omega_-)$  is minimal, by Theorem 2, it is Hurwitz stable. Moreover,  $G \sim (I, C_-, \Omega_-)$  satisfies Eq. (25). Therefore, according to Definition 4,  $G \sim (I, C_-, \Omega_-)$  is lossless bonded real.

(ii). Assume  $i\omega$  is not a pole of  $\Sigma(s)$ . Then the matrix  $i\omega I + i\Omega_{-}$  is invertible. Note that for all  $\operatorname{Re}[s] > 0$ ,

$$\Sigma(s) + \Sigma(s)^{\dagger}$$
(28)  
= Re [s]  $C_{-}(sI + i\Omega_{-})^{-1} \left( C_{-}(sI + i\Omega_{-})^{-1} \right)^{\dagger}$ .

By (28),  $\Sigma(i\omega) + \Sigma(i\omega)^{\dagger} = 0$ . Therefore, by Definition 5,  $\Sigma(s)$  is lossless positive real.

Finally, as a consequence of Eq. (24) between G(s) and  $\Sigma(s)$ , in the minimal realization case the properties (i) and (ii) are equivalent.  $\Box$ 

*Remark 3.* In fact, the relation between lossless bounded realness and lossless positive realness is well-known in electric networks theory, see. e.g., Anderson & Vongpanitlerd, 1973.

*Remark 4.* Here we have used the annihilation-operator form to study dissipative properties of quantum linear passive systems. Because the resulting matrices may be complex-valued, they can be viewed as the complex versions of lossless bounded real and lossless positive real in terms of the quadrature form, James, Nurdin & Petersen, 2008. In fact, if the quantum system is represented in the quadrature form, it is exactly the same lossless bounded real form as that in Anderson & Vongpanitlerd, 1973, Secs. 2.6 and 2.7 for classical linear systems.

### 4 Realizations for quantum linear passive systems

Several realizations of quantum linear passive systems are proposed in this section. The multi-inputmulti-output (MIMO) case is studied in Subsection 4.1. For the single-input-single-output (SISO) case, an independent-oscillator realization is proposed in Subsection 4.2.1, Fig. 2; a chain-mode realization is presented in Subsections 4.2.2, Fig. 3; and the uniqueness of the independent-oscillator realization is discussed in Subsection 4.2.3.

## 4.1 Realizations for multi-input-multi-output models

In this subsection a new realization for MIMO quantum linear passive systems is proposed.

Before presenting our realizations for quantum linear passive systems, we describe for completeness a realization proposed in Nurdin, 2010b and Petersen, 2011 using the series product to produce a realization of an noscillator system as a cascade of n one-oscillator systems.

We begin with the observation that every matrix  $n \times n$ matrix A admits a Schur decomposition  $A = U^{\dagger}A'U$ with U unitary and A' lower triangular. For a given quantum linear passive system  $G \sim (S, C_{-}, \Omega_{-})$ , we define a unitary transform  $a' \equiv Ua$ , such that  $A' = UAU^{\dagger}$  is lower triangular. Accordingly denote  $C' = C_{-}U^{\dagger}$  and  $\Omega' = U\Omega_{-}U^{\dagger}$ . The new system is thus  $G' \sim (S, C', \Omega')$ . A standard result from linear systems theory shows that the two systems G and G' have the same transfer function. In what follows we show the system G' has a cascade realization, Fig. 4.1. Because  $A' = -\frac{1}{2}C'^{\dagger}C' - i\Omega'$ is lower triangular, for j < k we have  $A'_{jk} = -\frac{1}{2}C'^{\dagger}C'_{k} - i\Omega'_{jk} = 0$ , so  $\Omega'_{jk} = \frac{i}{2}C'^{\dagger}C'_{k}$ . Therefore the lower triangular components are  $A'_{kj} = -\frac{1}{2}C'^{\dagger}C'_{j} - i\Omega'_{kj} = -\frac{1}{2}C'^{\dagger}C'_{j} - i\Omega'_{jk} \equiv -C'^{\dagger}_{k}C'_{j}$ .

Let us now set  $G_0 \sim (S, 0, 0)$  and  $G_k \sim (I, C'_k, \Omega'_{kk})$ then the new system G' has a the cascaded realization  $G' = G_n \triangleleft \cdots \triangleleft G_1 \triangleleft G_0$ , Fig. 4.1.



Fig. 1. A quantum linear passive system with n system oscillators is realised as a sequence of n components in series, each one having a one-mode oscillator.

Next we present a new realization for MIMO quantum linear passive systems, which may have: 1) a set of interconnected principal oscillators  $\tilde{a}_{\rm pr}$  that interact with the (possibly part of) environment  $\tilde{b}_{\rm pr}(t)$ ; 2) auxiliary oscillators  $\tilde{a}_{{\rm aux},1}$  and  $\tilde{a}_{{\rm aux},2}$ ) which only couple to the principal oscillators while otherwise being independent; 3) input-out channels  $\tilde{b}_{{\rm aux}}(t)$  that do not couple to the system oscillators.

**Theorem 4** A quantum linear passive system  $G = (I, C_{-}, \Omega_{-})$  can be unitarily transformed to another one

with the corresponding realization

$$d\tilde{a}_{\rm pr}(t) = -\left(\frac{\sigma(C_-)^2}{2} + i\tilde{\Omega}_1\right)\tilde{a}_{\rm pr}(t)dt - i\tilde{\Omega}_{21}\tilde{a}_{\rm aux,1}(t)dt$$

$$-i\Omega_{22}\tilde{a}_{\text{aux},2}(t)dt - \sigma(C_{-})dB_{\text{in,pr}}(t), \quad (29)$$

$$d\tilde{a}_{\mathrm{aux},1}(t) = -i\sigma(\Omega_3)\tilde{a}_{\mathrm{aux},1}(t)dt - i\Omega_{21}^{\dagger}\tilde{a}_{\mathrm{pr}}(t)dt, \qquad (30)$$

$$d\tilde{a}_{\mathrm{aux},2}(t) = -i\Omega_{22}^{\dagger}\tilde{a}_{\mathrm{pr}}(t)dt, \qquad (31)$$

$$dB_{\rm out,pr}(t) = \sigma(C_{-})\tilde{a}_{\rm pr}(t)dt + dB_{\rm in,pr}(t), \qquad (32)$$

$$dB_{\rm out,aux}(t) = dB_{\rm in,aux}(t), \tag{33}$$

where  $\tilde{\Omega}_1 = \tilde{\Omega}_1^{\dagger}$ ,  $\tilde{\Omega}_3 = \tilde{\Omega}_3^{\dagger}$ , and  $\sigma(X)$  denotes the diagonal matrix with diagonal entries being the non-zero singular values of the matrix X. Clearly, this new realization corresponds the a quantum linear passive system  $(I, \bar{C}, \bar{\Omega})$  with

$$\bar{C} \equiv \begin{bmatrix} \sigma(C_{-}) & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}, \quad \bar{\Omega} \equiv \begin{bmatrix} \tilde{\Omega}_{1} & \tilde{\Omega}_{21} & \tilde{\Omega}_{22} \\ \tilde{\Omega}_{21}^{\dagger} & \sigma(\tilde{\Omega}_{3}) & 0 \\ \tilde{\Omega}_{22}^{\dagger} & 0 & 0 \end{bmatrix}. \quad (34)$$

The proof is given in the Appendix.

By Proposition 1 and Theorems 1 and 4, we have the following result.

Corollary 2 For the realization (29)-(33),

- (1) the mode  $\tilde{a}_{pr}$  is both controllable and observable;
- (2) if the system  $G = (I, C_{-}, \Omega_{-})$  is Hurwitz stable, then  $\tilde{\Omega}_{21} \neq 0$  and  $\tilde{\Omega}_{22} \neq 0$ .

Remark 5. The realization (29)-(33) is in some sense like controllability and observability decomposition of quantum linear passive systems. For example, if the system is Hurwitz stable, then by Lemma 2, all the modes  $\tilde{a}_{\rm pr}, \tilde{a}_{\rm aux,1}$  and  $\tilde{a}_{\rm aux,2}$  are both controllable and observable. On the other hand, if  $\tilde{\Omega}_{21} = 0$  and  $\tilde{\Omega}_{22} = 0$ , then the modes  $\tilde{a}_{\rm aux,1}$  and  $\tilde{a}_{\rm aux,2}$  are neither controllable nor observable; in this case, the modes  $\tilde{a}_{\rm pr}$  span the controllable and observable subspace, while the modes  $\tilde{a}_{\rm aux,1}$ and  $\tilde{a}_{\rm aux,2}$  span the uncontrollable and unobservable subspace.

Remark 6. When m = 1, assuming minimal realization, from the proof given in the Appendix it can be seen that Theorem 4 reduces to Theorem 5 for the independent-oscillator realization of SISO systems to be discussed in Subsection 4.2.1.

#### 4.2 Realizations for single-input-single-output models

In this subsection, two realizations, namely the independent-oscillator realization and the chain-mode realization, of SISO linear passive systems are proposed.

4.2.1 Independent-oscillator realization



Fig. 2. The independent-oscillator realization: the principal mode is coupled to n-1 independent auxiliary modes. The principal mode couples to the field, while the auxiliary modes are independent other than that they couple to the principal mode.

Given a SISO quantum linear passive system  $G \sim (I, C_{-}, \Omega_{-})$  where

$$C_{-} = [\sqrt{\gamma_1} \dots \sqrt{\gamma_n}], \quad \Omega_{-} = (\omega_{jk})_{n \times n}, \tag{35}$$

we show how to find a unitarily equivalent realization in terms of a single oscillator (the coupling mode  $c_0$ , we also call it the principle mode) which is then coupled to n-1 auxiliary modes  $c_1, \dots, c_{n-1}$ . The auxiliary modes are themselves otherwise independent oscillators, Fig. 2.

**Theorem 5** There exists a unitary matrix T such that the transformed modes

$$\mathbf{c} = \begin{bmatrix} c_0 \\ c_1 \\ \vdots \\ c_{n-1} \end{bmatrix} \equiv T \mathbf{a}$$
(36)

have the following realizations

$$dc_0(t) = -(\gamma/2 + i\omega_0)c_0(t)dt$$
$$-\sum_{j=1}^{n-1} i\sqrt{\kappa_j}c_j(t) - \sqrt{\gamma}dB(t), \qquad (37)$$

$$dc_j(t) = -i\omega_j c_j(t)dt - i\sqrt{\kappa_j}c_0(t)dt, \qquad (38)$$

$$dB_{\rm out}(t) = \sqrt{\gamma}c_0(t) + dB(t), \tag{39}$$

where

$$\gamma \equiv \sum_{j=1}^{n} \gamma_j, \quad \omega_0 \equiv \frac{1}{\gamma} \sum_{j,k=1}^{n} \sqrt{\gamma_j \gamma_k} \omega_{jk}, \qquad (40)$$

and the other parameters  $\omega_j, \kappa_j \ (j = 1, ..., n - 1)$  are given in the proof.

**Proof.** Let R be a unitary matrix whose first row is  $R_{1j} = \sqrt{\gamma_j/\gamma}, (j = 1, ...n)$ . Set  $b'_j \equiv \sum_{k=1}^n R_{jk}a_k, j = 1, ...n$ . We have  $[b'_j, b'^*_k] = \delta_{jk}$ . Clearly  $L = C_- \mathbf{a} = \sqrt{\gamma}b'_1$  and  $[L, b'^*_j] = 0$  for j = 2, ...n. Let us apply a further unitary transformation V of the form V =

 $\begin{bmatrix} 1 & 0_{n-1}^{\mathsf{T}} \\ 0_{n-1} & \tilde{V} \end{bmatrix} \text{ with } 0_{n-1} \text{ the column vector of length} \\ n-1 \text{ with all zero entries and } \tilde{V} \text{ unitary in } \mathbb{C}^{(n-1)\times(n-1)} \\ \text{to be specified later. We set } \mathbf{c} = [c_0 \ c_1 \ \cdots \ c_{n-1}]^T \equiv \\ V \mathbf{b}' = VR \mathbf{a}. \text{ We have } L = \sqrt{\gamma}c_0. \text{ The Hamiltonian} \\ \text{takes the form } H = \mathbf{c}^{\dagger}VR\Omega R^{\dagger}V^{\dagger}\mathbf{c} = \mathbf{c}^{\dagger}\Omega'\mathbf{c}, \text{ where} \\ \Omega' \equiv \begin{bmatrix} 1 & 0_{n-1}^{\mathsf{T}} \\ 0_{n-1} & \tilde{V} \end{bmatrix} R\Omega_{-}R^{\dagger} \begin{bmatrix} 1 & 0_{n-1}^{\mathsf{T}} \\ 0_{n-1} & \tilde{V}^{\dagger} \end{bmatrix}. \text{ As the ma-$ 

trix V is still arbitrary except being unitary, we may choose it to diagonalize the lower right  $(n-1) \times (n-1)$ block of  $R\Omega R^{\dagger}$ , and with this choice we obtain  $\Omega'$  of

the form 
$$\Omega_{\rm IO} \equiv \begin{bmatrix} \omega_0 & \varepsilon_1^* \cdots & \varepsilon_{n-1}^* \\ \varepsilon_1 & \omega_1 & 0 \\ \vdots & \ddots & \\ \varepsilon_{n-1} & 0 & \omega_{n-1} \end{bmatrix}$$
. It can be readily

verified that  $\omega_0 = \frac{1}{\gamma} \sum_{jk=1}^n \sqrt{\gamma_j \gamma_k} \omega_{jk}$ . Set T = VR and the overall unitary transform is thus  $\mathbf{c} = T \mathbf{a}$ . Finally we may absorb the phases of the  $\varepsilon_k$  into the modes, so without loss of generality we may assume that they are real and non-negative, say  $\varepsilon_k \equiv \sqrt{\kappa_k}$ .  $\Box$ 

By Proposition 1 and Theorem 1, we have

**Corollary 3** For the realization (36)-(39) constructed in Proposition 5, if  $\kappa_j = 0$ , (j = 1, ..., n - 1,) then the mode  $c_i$  is neither controllable nor observable.

Because the two realizations,  $G \sim (I, C_{-}, \Omega_{-})$  with  $C_{-}, \Omega_{-}$  defined in (35) and that in (36)-(39), are unitarily equivalent, they have the same transfer function. In what follows we derive their transfer function.

The following lemma turns out to be useful.

Lemma 3 We have the algebraic identity that

$$\begin{pmatrix} \begin{bmatrix} a_0 & b_1 & \cdots & b_n \\ b_1 & a_1 & \ddots & 0 \\ \vdots & \ddots & \vdots \\ b_n & 0 & a_n \end{bmatrix}^{-1} \\ \end{pmatrix}_{\text{row 1,column 1}} = \frac{1}{a_0 - \sum_{k=1}^n \frac{b_k^2}{a_k}},$$

where  $(X)_{\text{row 1,column 1}}$  means the entry on the intersection of the first row and first column of a constant matrix X.

The proof is given in the Appendix.

We are now ready to present the transfer function.

**Corollary 4** The SISO quantum linear passive system  $G \sim (I, C_{-}, \Omega_{-})$  with  $C_{-}, \Omega_{-}$  defined in (35) has a transfer function of the form

$$G(s) = 1 - \frac{\gamma}{s + \frac{1}{2}\gamma + i\omega_0 + \sum_{k=1}^{n-1} \frac{\kappa_k}{s + i\omega_k}}.$$
 (41)

The proof follows Theorem 5 and Lemma 3.

Remark 7. Theorem 5 gives an independent-oscillator realization of a quantum linear passive system, Fig. 2. Unfortunately, because the unitary matrices V and R used in the proof of Theorem 5 are by no means unique, it is unclear whether this realization is unique or not, that is, whether the parameters  $\omega_i$  and  $\kappa_j$  are uniquely determined by the system parameters  $\gamma_i$  and  $\omega_{jk}$  in (35) or not. In Theorem 7 to be given in Subsection 4.2.3, we show that the independent-oscillator realization is unique under the assumption of minimal realization.

## 4.2.2 Chain-mode realization

In the subsection we present the chain-mode realization of SISO quantum linear passive systems.

Let  $G \sim (I, C_{\min}, \Omega_{\min})$  be a Hurwitz stable SISO quantum linear system with  $n_{\min}$  the number of system oscillators. We assume that  $\Omega_{\min}$  is diagonal and the entries of  $C_{\min}$  are non-negative; specifically,

$$\mathbf{\bar{a}} = \begin{bmatrix} \bar{a}_1 \\ \vdots \\ \bar{a}_{n_{\min}} \end{bmatrix}, \ \Omega_{\min} = \operatorname{diag}\left(\bar{\omega}_1, \cdots, \bar{\omega}_{n_{\min}}\right),$$
$$C_{\min} = \left[\sqrt{\bar{\gamma}_1}, \cdots, \sqrt{\bar{\gamma}_{n_{\min}}}\right].$$
(42)

Remark 8. Because the matrix  $\Omega_{\min}$  is Hermitian, it can always be diagonalized. Similarly by absorbing phases into system oscillators if necessary, the entries of the matrix  $C_{\min}$  can be taken to be non-negative. Thus, given a Hurwitz stable quantum linear passive system, one can always unitarily transform it to another one corresponding to (42). Moreover, by Proposition 4, minimality requires that  $\bar{\omega}_j \neq \bar{\omega}_k$  if  $j \neq k$ , and  $\bar{\gamma}_j \neq 0$ ,  $j = 1, \ldots, n_{\min}$ .

In what follows we unitarily transform the system  $G \sim (I, \Omega_{\min}, C_{\min})$  to a chain-mode realization of an assembly of interacting oscillators, Fig. 3.

**Theorem 6** For the system  $G \sim (I, C_{\min}, \Omega_{\min})$  defined by (42), there exists a unitary transform W such that the



Fig. 3. The Chain-mode realization: the principal mode is coupled to a non-damped mode which in turn is coupled to a finite chain of modes.

transformed modes

$$\begin{bmatrix} \tilde{c}_0 \\ \tilde{c}_1 \\ \vdots \\ \tilde{c}_{n_{\min}-1} \end{bmatrix} \equiv W \,\bar{\mathbf{a}}$$

have the following realization:

$$d\tilde{c}_{0}(t) = -(\bar{\gamma}/2 + i\tilde{\omega}_{0})\tilde{c}_{0}(t)dt -i\sqrt{\tilde{\kappa}_{1}}\tilde{c}_{1}(t)dt - \sqrt{\bar{\gamma}}dB(t), \qquad (43)$$
$$d\tilde{c}_{j}(t) = -i\tilde{\omega}_{j}\tilde{c}_{j}(t)dt - i\sqrt{\tilde{\kappa}_{j}}\tilde{c}_{j-1}(t)dt$$

$$-i\sqrt{\tilde{\kappa}_{j+1}}\tilde{c}_{j+1}(t)dt, \qquad (44)$$

$$d\tilde{c}_{n_{\min}-1}(t) = -i\tilde{\omega}_{n_{\min}-1}\tilde{c}_{n_{\min}-1}(t)dt$$

$$-i\sqrt{\tilde{\kappa}_{n_{\min}-1}}\tilde{c}_{n_{\min}-1}(t)dt \qquad (45)$$

$$dB_{\rm out}(t) = \sqrt{\bar{\gamma}}\tilde{c}_0(t)dt + dB(t), \tag{46}$$

where  $j = 1, ..., n_{\min} - 2$ , and the parameters  $\tilde{\omega}_j$  and  $\tilde{\kappa}_j$  are given respectively in (65) and (66) in the proof.

The proof is given in the Appendix.

*Remark 9.* In the literature of continued fraction, Gautschi, 2004; Hughes, Christ & Burghardt, 2009; Woods, et al., 2014, etc., the matrix

$$\mathcal{J} \equiv \begin{bmatrix} \tilde{\omega}_0 & \sqrt{\tilde{\kappa}_1} & 0 & \cdots & 0 & 0 \\ \sqrt{\tilde{\kappa}_1} & \tilde{\omega}_1 & \sqrt{\tilde{\kappa}_2} & 0 & 0 \\ 0 & \sqrt{\tilde{\kappa}_2} & \tilde{\omega}_2 & \ddots & & \vdots \\ \vdots & \ddots & \ddots & \sqrt{\tilde{\kappa}_{n_{\min}-2}} & 0 \\ 0 & 0 & \sqrt{\tilde{\kappa}_{n_{\min}-2}} & \sqrt{\tilde{\kappa}_{n_{\min}-1}} \\ 0 & 0 & \cdots & 0 & \sqrt{\tilde{\kappa}_{n_{\min}-1}} & \tilde{\omega}_{n_{\min}-1} \end{bmatrix}$$

is often called a *Jacobi matrix*. Clearly,  $\mathcal{J}$  is actually the

Hamiltonian matrix for the new system corresponding to the realization (43)-(46).

Because the two realizations,  $G \sim (I, C_{\min}, \Omega_{\min})$  defined by (42) and that in (43)-(46), are unitarily equivalent, they share the same transfer function. Next we study their transfer function.

## We begin with the following lemma.

#### Lemma 4 We have the algebraic identity that



where  $(X)_{row 1, column 1}$  means the entry on the intersection of the first row and first column of a matrix X.

The proof is given in the Appendix.

Based on Theorem 6 and Lemma 4, we may derive the transfer function.

**Corollary 5** The SISO quantum linear passive system  $G \sim (I, C_{\min}, \Omega_{\min})$  has a transfer function in the form of the continued fraction expansion in Eq. (47).

#### 4.2.3 Uniqueness of the independent-oscillator realization

In Subsection 4.2.1 an independent-oscillator realization for SISO quantum linear passive systems is proposed. From the construction it is unclear whether the parameters in this independent-oscillator realization are unique, Remark 7. In this subsection we show that they are indeed unique if minimality is assumed.

**Theorem 7** Given a minimal quantum linear passive system  $G \sim (I, C_{\min}, \Omega_{\min})$  in (42), its unitarily equivalent independent-oscillator realization is unique.

**Proof.** Firstly, for the minimal realization  $G \sim (I, C_{\min}, \Omega_{\min})$  in (42), by (40) and (67),  $\omega_0 = \tilde{\omega}_0$ . Secondly, by (41) and (47) we see the transfer function

takes the form

$$G(s) = 1 - \frac{\gamma}{s + \frac{\gamma}{2} + i\omega_0 + \Delta(s)},$$
(48)

where

-

$$\Delta(s) = \sum_{k=1}^{n_{\min}-1} \frac{\kappa_k}{s+i\omega_k}$$

$$= \frac{\tilde{\kappa}_1}{s+i\tilde{\omega}_1 + \frac{\tilde{\kappa}_2}{s+i\tilde{\omega}_2 + \cdot \cdot}} + \frac{\tilde{\kappa}_{n_{\min}-2}}{s+i\tilde{\omega}_{n_{\min}-2} + \frac{\tilde{\kappa}_{n_{\min}-1}}{s+\tilde{\omega}_{n_{\min}-1}}}$$
(49)
$$= \frac{\tilde{\kappa}_1}{(50)}$$

in the independent-oscillator and chain-mode realizations respectively. Replacing s with  $i\omega$  in (48), (49) and (50) we have

$$G(i\omega) = 1 + \frac{i\gamma}{\omega + \omega_0 - \frac{\gamma}{2}i - \hat{\Delta}(\omega)}$$

where

\_

$$\hat{\Delta}(\omega) \equiv i\Delta(i\omega) = \sum_{k=1}^{n_{\min}-1} \frac{\kappa_k}{\omega + \omega_k}$$
(51)

$$= \frac{\kappa_{1}}{\omega + \tilde{\omega}_{1} - \frac{\tilde{\kappa}_{2}}{\omega + \tilde{\omega}_{2} - \frac{\tilde{\kappa}_{n_{\min}-2}}{\omega + \tilde{\omega}_{n_{\min}-2} - \frac{\tilde{\kappa}_{n_{\min}-1}}{\omega + \tilde{\omega}_{n_{\min}-1}}}$$
(52)

in the independent-oscillator and chain-mode realizations respectively. By Theorem 6,  $\tilde{\omega}_j$  and  $\tilde{\kappa}_j$  in (52) are uniquely determined by  $C_{\min}$  and  $\Omega_{\min}$ , that is,  $\hat{\Delta}(\omega)$  is unique. On the other hand, because  $G = (I, C_{\min}, \Omega_{\min})$ is minimal, in (51)  $\omega_j \neq \omega_k$  if  $j \neq k$ , and  $\kappa_i \neq 0$ . Clearly, for this single pole fraction form of  $\hat{\Delta}(\omega)$  in (51),  $\kappa_k$  and  $\omega_k$  are unique.  $\Box$ 

We notice that (48) implies that

$$\Sigma\left(s\right) = \frac{1}{2} \frac{\gamma}{s + i\omega_0 + \Delta\left(s\right)} \tag{53}$$

with  $\Delta(s)$  given by (50).

$$G(s) = I - \frac{\bar{\gamma}}{s + \frac{1}{2}\bar{\gamma} + i\omega_0 + \frac{\tilde{\kappa}_1}{s + i\tilde{\omega}_1 + \ddots}} + \frac{\tilde{\kappa}_{n_{\min}-2}}{s + i\tilde{\omega}_{n_{\min}-2} + \frac{\tilde{\kappa}_{n_{\min}-1}}{s + i\tilde{\omega}_{n_{\min}-1}}}.$$
(47)

Remark 10. Given  $\Delta(s)$  in (49) and (50), by (53) an explicit form of  $\Sigma(s)$  can be constructed, subsequently a quantum linear passive system  $G(s) = (I - \Sigma(s))((I + \Sigma(s)))^{-1}$  can be constructed. According to (48), G(s) constructed in this way is always a genuine quantum system.

## 5 Conclusion

In this paper we have studied the realization theory of quantum linear systems. We have shown the equivalence between controllability and observability of general quantum linear systems, and in particular in the passive case they are equivalent to Hurwitz stability. Based on controllability and observability, a special form of realization has been proposed for general quantum linear systems which can be regarded as the complex-domain counterpart of the so-called decoherence-free subsystem decomposition studied in Yamamoto, 2014. Specific to quantum linear passive systems, formulas for calculating the cardinality of minimal realizations are proposed. A specific realization is proposed for the multi-input-multioutput case which is related to controllability and observability decomposition. Finally, two realizations, the independent-oscillator realization and the chain-mode realization, have been derived for the single-input-singleoutput case. It is expected that these results will find applications in quantum systems design.

## Acknowledgment

The authors wish to thank Daniel Burgarth for pointing out Reference Woods, et al., 2014. The second author would like to thank Runze Cai, Lei Cui and Zhiyang Dong for helpful discussions.

#### Appendix.

Proof of Lemma 1. We first show that the dimension of the space  $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$  is even. If a nonzero vector

$$v = \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} \in \operatorname{Ker}\left(\mathbf{O}_s\right) \cap \operatorname{Ker}\left(\mathbf{O}_s J_n\right)$$
(54)

with 
$$v_1, v_2 \in \mathbb{C}^n$$
, then  $\mathcal{C}\begin{bmatrix}v_1\\v_2\end{bmatrix} = \mathcal{C}J_n\begin{bmatrix}v_1\\v_2\end{bmatrix} = 0$ . Actually,

$$\mathcal{C}\begin{bmatrix}v_1\\v_2\end{bmatrix} = \mathcal{C}J_n\begin{bmatrix}v_1\\v_2\end{bmatrix} = 0 \iff \mathcal{C}\begin{bmatrix}v_1 & 0\\0 & v_2\end{bmatrix} = 0.$$
(55)

On the other other hand, by (54) we also have  $\mathcal{C}J_n\Omega\begin{bmatrix}v_1\\v_2\end{bmatrix}=0, \quad \mathcal{C}J_n\Omega J_n\begin{bmatrix}v_1\\v_2\end{bmatrix}=0, \text{ which are equiv-}$ alent to  $\mathcal{C}J_n\Omega\begin{bmatrix}v_1 & 0\\ 0 & v_2\end{bmatrix}=0.$  So we have

$$\mathcal{C}J_n\Omega\begin{bmatrix}v_1\\v_2\end{bmatrix} = 0, \quad \mathcal{C}J_n\Omega J_n\begin{bmatrix}v_1\\v_2\end{bmatrix} = 0$$
$$\iff \mathcal{C}J_n\Omega\begin{bmatrix}v_1\\0\end{bmatrix} = 0, \quad \mathcal{C}J_n\Omega\begin{bmatrix}0\\v_2\end{bmatrix} = 0. \tag{56}$$

Analogously it can be shown that

$$\mathcal{C}(J_n\Omega)^k \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = 0, \quad \mathcal{C}J_n(J_n\Omega)^k \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = 0 \quad (57)$$
$$\iff \mathcal{C}(J_n\Omega)^k \begin{bmatrix} v_1 \\ 0 \end{bmatrix} = 0, \quad \mathcal{C}(J_n\Omega)^k \begin{bmatrix} 0 \\ v_2 \end{bmatrix} = 0, \quad k \ge 1.$$

(55), (56) and (57) indicate that

$$v = \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} \in \operatorname{Ker} \left( \mathbf{O}_s \right) \cap \operatorname{Ker} \left( \mathbf{O}_s \mathbf{J}_n \right)$$
$$\iff \begin{bmatrix} v_1 \\ 0 \end{bmatrix}, \begin{bmatrix} 0 \\ v_2 \end{bmatrix} \in \operatorname{Ker} \left( \mathbf{O}_s \right) \cap \operatorname{Ker} \left( \mathbf{O}_s J_n \right).$$

As a result, one can choose an orthonormal basis of  $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$  to be one of the form  $\begin{bmatrix} v_1 \\ 0 \end{bmatrix}, \begin{bmatrix} 0 \\ v_1^{\#} \end{bmatrix}, \cdots, \begin{bmatrix} v_l \\ 0 \end{bmatrix}, \begin{bmatrix} 0 \\ v_l^{\#} \end{bmatrix}$ . Therefore, the di-

mension of the space  $\operatorname{Ker}(\mathbf{O}_s) \cap \operatorname{Ker}(\mathbf{O}_s J_n)$  is even. Here we take it to be 2l.

Secondly, we construct  $V_1 \in \mathbb{C}^{2n \times 2l}$ . Noticing

$$\operatorname{Ker}\left(\mathbf{O}_{s}J_{n}[I_{n} \ 0_{n}]^{T}\right) = \operatorname{Ker}\left(\mathbf{O}_{s}[I_{n} \ 0_{n}]^{T}\right),$$

we have

$$\begin{bmatrix} v_i \\ 0 \end{bmatrix} \in \operatorname{Ker}\left(\mathbf{O}_s\right) \cap \operatorname{Ker}\left(\mathbf{O}_s J_n\right) \Longleftrightarrow v_i \in \operatorname{Ker}\left(\mathbf{O}_s \begin{bmatrix} I_n \\ 0_n \end{bmatrix}\right)$$

Thus it is sufficient to construct the orthonormal basis vectors  $v_1, \ldots, v_l$  for the space  $\operatorname{Ker}\left(\mathbf{O}_s[I_n \ 0_n]^T\right)$ . This can be done by the Gram-Schmidt orthogonalisation procedure. Define  $V_1 \equiv \begin{bmatrix} v_1 \cdots v_l & 0 \cdots & 0 \\ 0 \cdots & 0 & v_1^{\#} \cdots & v_l^{\#} \end{bmatrix} \in \mathbb{C}^{2n \times 2l}$ . For the above construction,  $\operatorname{Range}(V_1) = \operatorname{Ker}\left(\mathbf{O}_s\right) \cap \operatorname{Ker}\left(\mathbf{O}_s J_n\right)$ . (8) is established.

Thirdly, we construct the matrix  $V_2$ . If a normalized vector  $v_{l+1} \in \mathbb{C}^n$  such that for all  $k = 1, \ldots, l$ ,  $v_{l+1}^{\dagger}v_k = 0$ , then  $(v_{l+1}^{\#})^{\dagger}v_k^{\#} = 0$ . That is, the normalized vectors  $\begin{bmatrix} v_{l+1} \\ 0 \end{bmatrix}$  and  $\begin{bmatrix} 0 \\ v_{l+1}^{\#} \end{bmatrix}$  are orthogonal to the

space Range( $V_1$ ). Of course  $\begin{bmatrix} v_{l+1} \\ 0 \end{bmatrix}$  and  $\begin{bmatrix} 0 \\ v_{l+1}^{\#} \end{bmatrix}$  are

orthogonal to each other too. By the Gram-Schmidt orthogonalisation procedure an orthonormal basis  $\{v_{l+1}, \ldots, v_n\}$  can be found for the orthogonal space of the space spanned by the vectors  $\{v_1, \ldots, v_l\}$ . The an orthonormal matrix  $V_2$  can be constructed to be

$$V_2 \equiv \begin{bmatrix} v_{l+1} \cdots v_n & 0 & \cdots & 0 \\ 0 & \cdots & 0 & v_{l+1}^{\#} & \cdots & v_n^{\#} \end{bmatrix} \in \mathbb{C}^{2n \times 2(n-l)}.$$

Fourthly, define  $V \equiv [V_1 \ V_2]$ . Clearly,  $V^{\dagger}V = I_{2n}$  which establishes (9).

Finally, because  $V_1^{\dagger} J_n = J_l V_1^{\dagger}$ , we have

$$V^{\dagger}J_{n}V = \begin{bmatrix} V_{1}^{\dagger}J_{n}V_{1} & V_{1}^{\dagger}J_{n}V_{2} \\ V_{2}^{\dagger}J_{n}V_{1} & V_{2}^{\dagger}J_{n}V_{2} \end{bmatrix} = \begin{bmatrix} J_{l} & 0 \\ 0 & J_{n-l} \end{bmatrix},$$

which is (10).  $\Box$ 

Proof of Proposition 4. Without loss of generality, assume that  $\Omega_{-}$  is diagonal. (Otherwise, there exists a unitary matrix T such that  $\overline{\Omega} = T\Omega_{-}T^{\dagger}$  is diagonal. Correspondingly, denote  $\overline{P}_{\omega} = TP_{\omega}T^{\dagger}$  and  $\overline{C} = C_{-}T^{\dagger}$ . Then  $\bar{C}\bar{P}_{\omega}\bar{C}^{\dagger} = C_{-}P_{\omega}C^{\dagger}$ .) Let there be r non-zero entries in the row vector  $C_{-}$ . Because  $\Omega_{-}$  is diagonal, if the *i*th element of  $C_{-}$  is zero, then the *i*th column of the matrix in (7) is a zero column. As a result, for minimality we need only consider non-zero elements of  $C_{-}$ . Without loss of generality, assume  $C_{-} = [C_{1} \ 0]$ , where  $C_{1} = [c_{1} \ c_{2} \ \cdots \ c_{r}]$  with  $c_{i} \neq 0$ ,  $(i = 1, \ldots, r)$ . Correspond- $[\Omega_{1} \ 0]$ 

ingly, partition  $\Omega_{-}$  as  $\Omega_{-} = \begin{bmatrix} \Omega_{1} & 0 \\ 0 & \Omega_{2} \end{bmatrix}$ , where  $\Omega_{1}$  is a  $r \times r$  square diagonal matrix with  $\omega_{1}, \ldots, \omega_{r}$  being diagonal

entries. Clearly, rank 
$$\begin{bmatrix} C_{-} \\ C_{-}\Omega_{-} \\ \vdots \\ C_{-}\Omega_{-}^{n-1} \end{bmatrix} = \operatorname{rank} \begin{bmatrix} C_{1} \\ C_{1}\Omega_{1} \\ \vdots \\ C_{1}\Omega_{1}^{r-1} \end{bmatrix}.$$

Notice that

$$\begin{bmatrix} C_1 \\ C_1\Omega_1 \\ \vdots \\ C_1\Omega_1^{r-1} \end{bmatrix} = \begin{bmatrix} 1 & 1 & \cdots & 1 \\ \omega_1 & \omega_2 & \cdots & \omega_r \\ \vdots & \vdots & \vdots & \vdots \\ \omega_1^{r-1} & \omega_2^{r-1} & \cdots & \omega_r^{r-1} \end{bmatrix} \begin{bmatrix} c_1 \\ c_2 \\ \vdots \\ c_r \end{bmatrix}.$$

According to Lemma 2 and noticing  $c_i \neq 0$  for  $i = 1, \ldots, r$ ,

$$n_{\min} = \operatorname{rank} \begin{bmatrix} C_1 \\ C_1 \Omega_1 \\ \vdots \\ C_1 \Omega_1^{r-1} \end{bmatrix} = \operatorname{rank} \begin{bmatrix} 1 & 1 & \cdots & 1 \\ \omega_1 & \omega_2 & \cdots & \omega_r \\ \vdots & \vdots & \vdots & \vdots \\ \omega_1^{r-1} & \omega_2^{r-1} & \cdots & \omega_r^{r-1} \end{bmatrix}$$

Let  $\ell$  be the total number of distinct diagonal entries of the matrix  $\Omega_1$ . By a property of the Vandermonde matrices,  $\ell = n_{\min}$ . Finally, denote the distinct eigenvalues of  $\Omega_1$  by  $\hat{\omega}_1, \ldots, \hat{\omega}_\ell$ . For each  $i = 1, \ldots, \ell$ , because  $c_i \neq 0, C_- P \hat{\omega}_i C_-^{\dagger} \neq 0$ . So we have shown that the number  $n_{\min}$  of system oscillators of a minimal realization  $(S, C_{\min}, \Omega_{\min})$  equals the total number of elements of the set  $\sigma(\Omega_-, C_-)$  defined in (27).  $\Box$ 

Proof of Theorem 4. The proof can be done by construction. Let rank $(C_{-}) = r > 0$ . Firstly, according toBernstein, 2009, Theorem 5.6.4 there exist unitary matrices  $R_1 \in \mathbb{C}^{m \times m}$  and  $R_2 \in \mathbb{C}^{n \times n}$  such that  $R_1C_-R_2 = \begin{bmatrix} \sigma(C_-)_{r \times r} & 0 \\ 0 & 0 \end{bmatrix}$  where  $\sigma(C_-)$  is a diagonal matrix with

diagonal entries being singular values of the matrix  $C_{-}$ . Partition the matrix  $R_{2}^{\dagger}\Omega_{-}R_{2}$  accordingly, and denote

$$\bar{\Omega} = \begin{bmatrix} \tilde{\Omega}_1 & \tilde{\Omega}_2 \\ \tilde{\Omega}_2^{\dagger} & \tilde{\Omega}_3 \end{bmatrix} \equiv R_2^{\dagger} \Omega_- R_2. \text{ Define the unitary trans-}$$

formations  $\begin{bmatrix} \tilde{b}_{in,pr}(t) \\ \tilde{b}_{in,aux}(t) \end{bmatrix} \equiv R_1 b(t), \begin{bmatrix} \tilde{b}_{out,pr}(t) \\ \tilde{b}_{out,aux}(t) \end{bmatrix} \equiv R_1 b_{out}(t), \begin{bmatrix} \tilde{a}_{pr}(t) \\ a_{aux}(t) \end{bmatrix} \equiv R_2^{\dagger} a(t)$ , where all the first blocks

on the left-hand side are a row vector of dimension r. Then G is unitarily equivalent to the following system

$$\dot{\tilde{a}}_{\rm pr} = -(\sigma(C_-)^2/2 + i\tilde{\Omega}_1)\tilde{a}_{\rm pr} - i\tilde{\Omega}_2 a_{\rm aux} -\sigma(C_-)\tilde{b}_{\rm in,pr}(t),$$
(58)

$$\tilde{u}_{aux} = -i \Omega_2 u_{pr} - i \Omega_3 u_{aux}, \tag{39}$$

$$\tilde{i}_{aux} = -i \Omega_2 u_{pr} - i \Omega_3 u_{aux}, \tag{39}$$

$$b_{\text{out,pr}} = \sigma(C_{-})a_{\text{pr}} + b_{\text{in,pr}}(t), \tag{60}$$

$$b_{\rm out,aux} = b_{\rm in,aux}(t). \tag{61}$$

By Schur decomposition there exists a unitary matrix

$$T \in \mathbb{C}^{(n-r) \times (n-r)}$$
 such that  $\tilde{\Omega}_3 = T \begin{bmatrix} \sigma(\Omega_3) & 0 \\ 0 & 0 \end{bmatrix} T^{\dagger}$ . Ac-

cordingly, denote  $[\Omega_{21} \ \Omega_{22}] \equiv \Omega_2 T^{\dagger}$ . As a result, applying the unitary transformation

$$\begin{bmatrix} \tilde{a}_{\rm pr} \\ \tilde{a}_{\rm aux,1} \\ \tilde{a}_{\rm aux,2} \end{bmatrix} \equiv \begin{bmatrix} I_{r \times r} & 0 \\ 0 & T \end{bmatrix} \begin{bmatrix} \tilde{a}_{\rm pr} \\ a_{\rm aux} \end{bmatrix}$$
(62)

to (58)-(59) yields the final realization (29)-(33). Clearly the realization (29)-(33) corresponds to a quantum linear passive system with parameters given in (34).

Proof of Lemma 3. We show this by induction. It is clear true for n = 1, so we the assume it is true for a given nand establish for n+1. Let us write  $E_{11}(M)$  for the first entry (row 1, column 1) of a matrix M. Let us consider a sequence

$$M_{n} = \begin{bmatrix} a_{0} \ b_{1} \ \cdots \ b_{n} \\ b_{1} \ a_{1} \ \ddots \ 0 \\ \vdots \ \ddots \\ b_{n} \ 0 \ a_{n} \end{bmatrix}$$

of matrices, then  $M_{n+1} \equiv \begin{bmatrix} M_n & b_{n+1}e_n \\ b_{n+1}e_n^\top & a_{n+1} \end{bmatrix}$ , where

 $e_n = [1 \ 0 \ \cdots 0]^T \in \mathbb{C}^{n+1}$ . We recall the Schur-Feshbach inversion formula for a matrix in block form

$$\begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix}^{-1} = \begin{bmatrix} Y^{-1} & -Y^{-1}A_{12}A_{22}^{-1} \\ -A_{22}^{-1}A_{21}Y^{-1} & A_{22}^{-1} + A_{22}^{-1}A_{21}Y^{-1}A_{12}A_{22}^{-2} \end{bmatrix}$$
(63)

where  $Y = A_{11} - A_{21}A_{22}^{-1}A_{21}$ . From the Schur-Feshbach formula we deduce that  $E_{11}(M_{n+1}^{-1}) =$  $E_{11}\left(\left(M_n - \frac{b_{n+1}^2}{a_{n+1}}e_ne_n^{\top}\right)^{-1}\right).$  However, the matrix  $M_n - (b_{n+1}^2/a_{n+1})e_ne_n^{\top}$  is identical to  $M_n$  except that we replace the first row first column entry  $a_0$  with  $a_0 - (b_{n+1}^2/a_{n+1})$ , and by assumption we should then have

$$E_{11}\left((M_n - \frac{b_{n+1}^2}{a_{n+1}}e_ne_n^{\top})^{-1}\right) = \frac{1}{\left(a_0 - \frac{b_{n+1}^2}{a_{n+1}}\right) - \sum_{k=1}^n \frac{b_k^2}{a_k}}$$

This establishes the formula for n+1, and so the formula is true by induction.  $\Box$ 

*Proof of Proposition 6.* The spectral distribution  $\Phi$  associated with a SISO system  $G \sim (S, C_{-}, \Omega_{-})$  is defined through the Stieltjes' integral, i.e.,  $\int_{-\infty}^{\infty} e^{it\omega} d\Phi(\omega) =$  $\frac{1}{C_{-}C_{-}^{\dagger}}C_{-}e^{it\Omega_{-}}C_{-}^{\dagger}$ , where the normalization coefficient  $C_{-}C_{-}^{\dagger} > 0$ . In particular, in terms of the specific minimal realization  $G \sim (S, C_{\min}, \Omega_{\min})$  given in (42), we have

$$d\Phi\left(\omega\right) = \sum_{j=1}^{n_{\min}} \frac{\bar{\gamma}_j}{\bar{\gamma}} \delta\left(\omega - \bar{\omega}_j\right) \, d\omega \equiv \bar{\mu}(\omega) d\omega, \qquad (64)$$

where  $\bar{\gamma} \equiv \sum_{j=1}^{n_{\min}} \bar{\gamma}_j$ . That is, the cardinality of the support of  $d\Phi$  is exactly the number of oscillators  $n_{\min}$  in the minimal realization of  $G \sim (S, C_{\min}, \Omega_{\min})$ . The spectral distribution defined in (64) has only finitely many point supports. We define an inner product for polynomials in the field of real numbers in terms of this discrete spectral distribution. More specifically, given two real polynomials  $P(\omega)$  and  $Q(\omega)$ , define their inner product with respect to  $\bar{\mu}$  to be

$$\langle P, Q \rangle_{\bar{\mu}} \equiv \int_{-\infty}^{\infty} P(\omega)Q(\omega)\bar{\mu}(\omega)d\omega = \sum_{j=1}^{n_{\min}} \frac{\bar{\gamma}_j}{\bar{\gamma}} P(\bar{\omega}_j)Q(\bar{\omega}_j).$$

The norm of a polynomial  $P(\omega)$  is of course  $||P|| \equiv$  $\sqrt{\langle P, P \rangle_{\bar{\mu}}}$ . Next we introduce a sequence of  $n_{\min}$  orthogonal polynomials  $\{P_i\}$ , which are defined via the Gram-Schmidt orthogonalization:

$$P_0(\omega) \equiv 1, \ P_j(\omega) = \omega^j - \sum_{k=0}^{j-1} \frac{\langle \omega^j, P_k \rangle_{\bar{\mu}}}{\langle P_k, P_k \rangle_{\bar{\mu}}} P_k(\omega),$$

where  $j = 1, \ldots, n_{\min} - 1$ ,  $\langle \omega^j, P_k \rangle_{\bar{\mu}}$  is to be under-stood as  $\langle \omega^j, P_k \rangle_{\bar{\mu}} = \int_{-\infty}^{\infty} \omega^j P_k(\omega) \bar{\mu}(\omega) d\omega$ . It is easy to verify that the above orthogonal polynomial sequence  $\{P_j\}_{j=0}^{n_{\min}}$  satisfies the following three-term recurrence relation, Gautschi, 2004, Theorem 1.27

$$P_{k+1}(\omega) = (\omega - \tilde{\omega}_k)P_k(\omega) - \sqrt{\tilde{\kappa}_k}P_{k-1}(\omega),$$

where  $k = 0, ..., n_{\min} - 1$ ,  $\tilde{\kappa}_0 \equiv ||P_0||$  and the convention  $P_{-1} \equiv 0$  is assumed. Clearly,

$$\tilde{\omega}_k = \frac{\langle \omega P_k, P_k \rangle_{\bar{\mu}}}{\langle P_k, P_k \rangle_{\bar{\mu}}}, \quad k = 0, \dots, n_{\min} - 1, \tag{65}$$

and

$$\tilde{\kappa}_k = \sqrt{\frac{\langle P_k, P_k \rangle_{\bar{\mu}}}{\langle P_{k-1}, P_{k-1} \rangle_{\bar{\mu}}}}, \quad k = 1, \dots, n_{\min} - 1.$$
(66)

(Note that  $\tilde{\kappa}_k \neq 0, \ k = 0, \dots, n_{\min} - 1$ .) According to (65), we have

$$\tilde{\omega}_0 = \frac{1}{\bar{\gamma}} \sum_{j=1}^{n_{\min}} \bar{\gamma}_j \bar{\omega}_j.$$
(67)

By normalizing  $\{P_j\}_{j=0}^{n_{\min}}$ , that is define  $\tilde{P}_j \equiv \frac{1}{\|P_j\|}P_j$ , we can get a set of orthonormal polynomial sequence  $\{\tilde{P}_j\}_{j=0}^{n_{\min}}$ . We define a new set of oscillators to be

$$\tilde{c}_0 \equiv \sum_{j=1}^{n_{\min}} \sqrt{\frac{\bar{\gamma}_j}{\bar{\gamma}}} \tilde{P}_0(\bar{\omega}_j) \bar{a}_j, \tag{68}$$

$$\tilde{c}_k \equiv \sum_{j=1}^{n_{\min}} \sqrt{\frac{\bar{\gamma}_j}{\bar{\gamma}}} \tilde{P}_k(\bar{\omega}_j) \bar{a}_j, \quad k = 1, \dots, n_{\min} - 1.$$
(69)

It can be verified that the transformation (68)-(69) is unitary. Moreover,

$$\tilde{c}_0 = \frac{1}{\sqrt{\bar{\gamma}}} \sum_{j=1}^{n_{\min}} \sqrt{\bar{\gamma}_j} \bar{a}_j, \qquad (70)$$

and the canonical commutation relations  $[\tilde{c}_0, \tilde{c}_k] = [\tilde{c}_0, \tilde{c}_k^*] = 0$ ,  $[\tilde{c}_j, \tilde{c}_k^*] = \delta_{jk}$  for  $j, k = 1, \ldots, n_{\min} - 1$ . By (70),

$$\tilde{L} = \sqrt{\bar{\gamma}}\tilde{c}_0. \tag{71}$$

Define matrices

$$Q = \begin{bmatrix} \tilde{P}_0(\bar{\omega}_1) & \cdots & \tilde{P}_0(\bar{\omega}_{n_{\min}}) \\ \vdots & \ddots & \vdots \\ \tilde{P}_{n_{\min}-1}(\bar{\omega}_1) & \cdots & \tilde{P}_{n_{\min}-1}(\bar{\omega}_{n_{\min}}) \end{bmatrix} \equiv \begin{bmatrix} \tilde{P}_0(\bar{\omega}) \\ \vdots \\ \tilde{P}_{n_{\min}-1}(\bar{\omega}) \\ (72) \end{bmatrix}$$

and  $\Gamma \equiv \text{diag}\left(\sqrt{\frac{\bar{\gamma}_1}{\bar{\gamma}}}, \cdots, \sqrt{\frac{\bar{\gamma}_{n_{\min}}}{\bar{\gamma}}}\right)$ . It can be shown that

$$\sum_{k=0}^{n_{\min}-1} \frac{\bar{\gamma}_i}{\bar{\gamma}} \tilde{P}_k(\bar{\omega}_i) \tilde{P}_k(\bar{\omega}_j) = \delta_{ij}, \quad i, j = 1, \dots, n_{\min}, \quad (73)$$

see, e.g., Gautschi, 2004, Eq. (1.1.14). By (73), it can be verified that the inverse matrix of the matrix Q turns out to be  $Q^{-1} = \Gamma^2[\tilde{P}_0(\bar{\omega})^{\dagger} \dots, \tilde{P}_{n_{\min}-1}(\bar{\omega})^{\dagger}]$ . Thus we have

$$\begin{bmatrix} \tilde{c}_0 \\ \tilde{c}_1 \\ \vdots \\ \tilde{c}_{n_{\min}-1} \end{bmatrix} = Q\Gamma \begin{bmatrix} \bar{a}_1 \\ \bar{a}_2 \\ \vdots \\ \bar{a}_{n_{\min}} \end{bmatrix}.$$

With this, the Hamiltonian of the minimal realization can be re-written as

$$\sum_{j=1}^{n_{\min}} \bar{\omega}_j \bar{a}_j^* \bar{a}_j = \begin{bmatrix} \tilde{c}_0 \\ \tilde{c}_1 \\ \vdots \\ \tilde{c}_{n_{\min}-1} \end{bmatrix}^{\dagger} \tilde{H} \begin{bmatrix} \tilde{c}_0 \\ \tilde{c}_1 \\ \vdots \\ \tilde{c}_{n_{\min}-1} \end{bmatrix}$$

where, according to (72) and (73), the new Hamiltonian matrix is

$$\tilde{H} = \begin{bmatrix} \tilde{\omega}_{0} & \sqrt{\tilde{\kappa}_{1}} & 0 & \cdots & 0 & 0 \\ \sqrt{\tilde{\kappa}_{1}} & \tilde{\omega}_{1} & \sqrt{\tilde{\kappa}_{2}} & 0 & 0 \\ 0 & \sqrt{\tilde{\kappa}_{2}} & \tilde{\omega}_{2} & \ddots & & \vdots \\ \vdots & \ddots & \ddots & \sqrt{\tilde{\kappa}_{n_{\min}-2}} & 0 \\ 0 & 0 & \sqrt{\tilde{\kappa}_{n_{\min}-2}} & \tilde{\omega}_{n_{\min}-2} & \sqrt{\tilde{\kappa}_{n_{\min}-1}} \\ 0 & 0 & \cdots & 0 & \sqrt{\tilde{\kappa}_{n_{\min}-1}} & \tilde{\omega}_{n_{\min}-1} \end{bmatrix}$$

With the new coupling operator  $\tilde{J}$  defined (71) and new Hamiltonian matrix  $\tilde{H}$  defined above, the realization (43)-(46) can be obtained.  $\Box$ 

*Proof of Lemma 4.* We again use induction. The formula is clearly true for n = 1. Let us set  $N_n = \begin{bmatrix} a_0 & b_1 & 0 \end{bmatrix}$ 

$$\begin{vmatrix} b_1 & a_1 & \ddots \\ & \ddots & \ddots & b_n \\ 0 & b_n & a_n \end{vmatrix} \text{ and so } N_{n+1} = \begin{bmatrix} N_n & b_{n+1}f_n \\ b_{n+1}f_n^\top & a_{n+1} \end{bmatrix}$$

where  $f_n = [0 \cdots 0 \ 1]^T \in \mathbb{C}^{n+1}$ . Let us write  $E_{11}(M)$ for the first entry (row 1, column 1) of a matrix M. We deduce from the Schur-Feshbach formula (63) that  $E_{11}(N_{n+1}^{-1}) = E_{11}\left((N_n - \frac{b_{n+1}^2}{a_{n+1}}f_nf_n^{\top})^{-1}\right)$ . However, the matrix  $N_n - (b_{n+1}^2/a_{n+1})f_nf_n^{\top}$  is identical to  $N_n$ except that we replace the last row, last column entry  $a_n$  with  $a_n - (b_{n+1}^2/a_{n+1})$ , and if by assumption the relation is true for n we deduce the formula for n+1. The formula is true by induction.  $\Box$ 

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